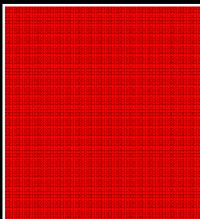


2012

**WURTH ELECTRONICS**

## EMC Seminar

Introduction to Concepts and  
Techniques TORONTO



# Agenda

Magnetic Materials / Inductors

Common Mode Filters

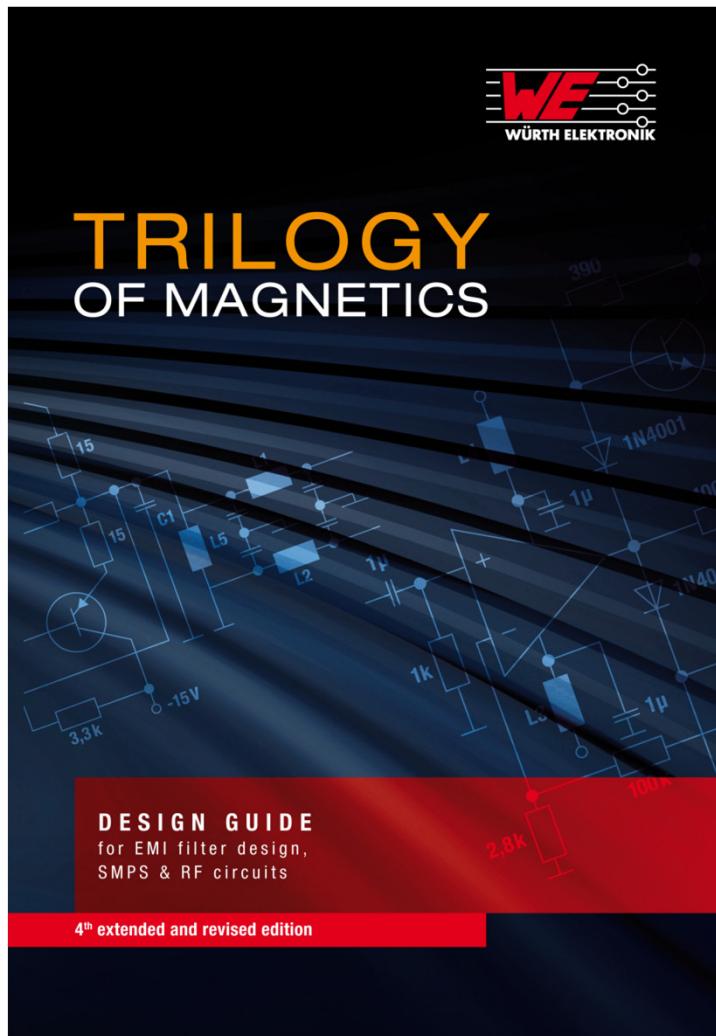
Chip Bead Ferrites

Shields

# TRILOGY OF MAGNETICS



## Supplement Material to Today's Seminar



- 1. Basic Principles of Magnetics
- 2. Magnetics Components
- 3. Filter Circuits
- 4. Applications

→ concrete examples on more than 300 pages

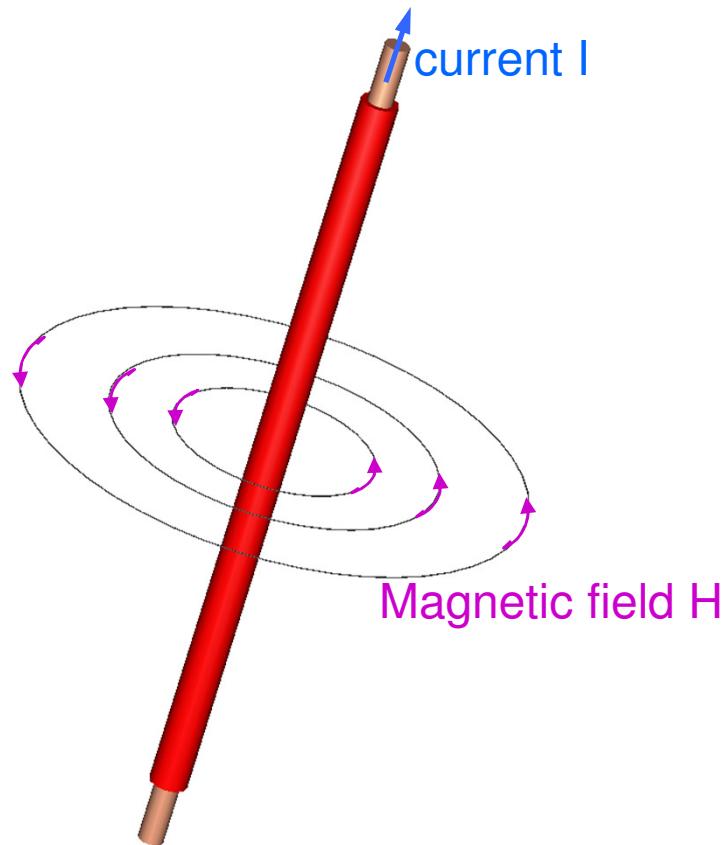


# MAGNETIC MATERIALS AND INDUCTORS

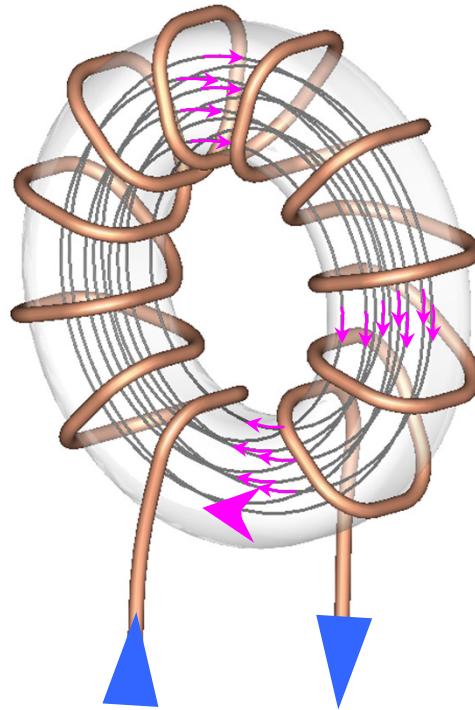
# The Magnetic Field (H)

When current flows through a wire it generates a magnetic field

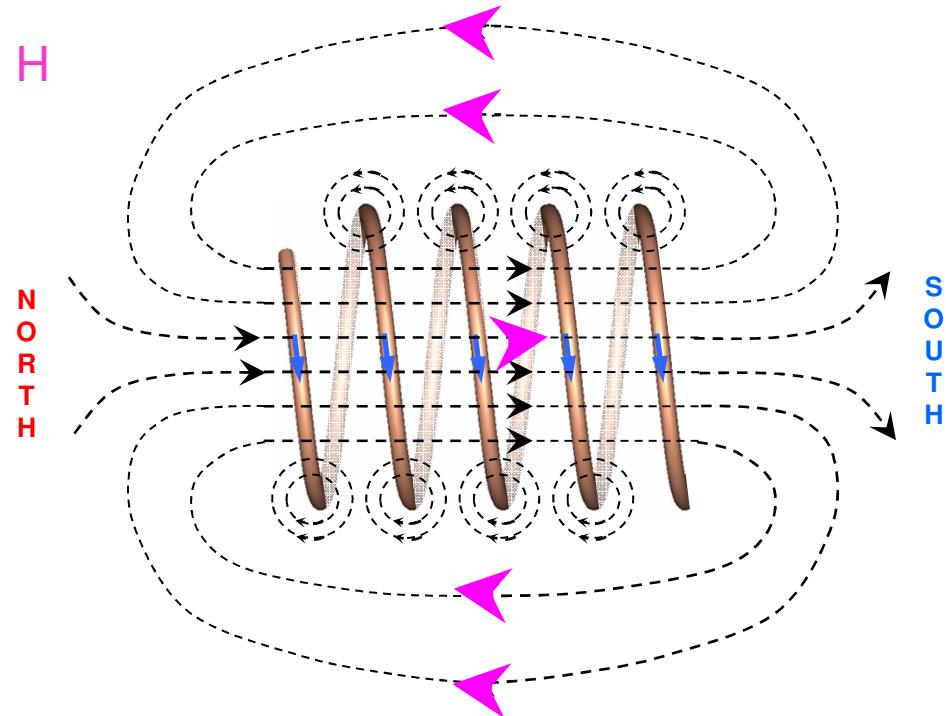
Field model



# The Inductor and its Magnetic Field



Magnetic field  $H$



Current  $I$

Many EMC solutions consist of **inductive** components

## Materials Used to Make Inductors



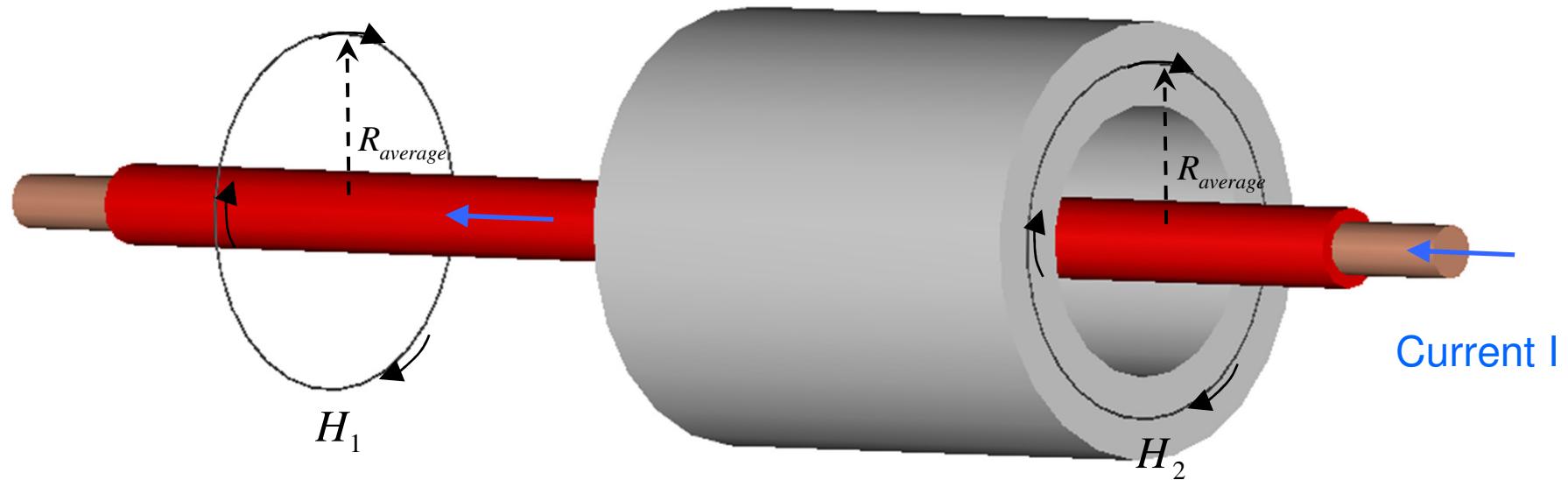
Iron Alloys

Fe + Other  
materials

Ferrites

Ni + Zn  
Mn

# Magnetic H and B Fields



$$H_1 = H_2 = H = \frac{I}{2 \cdot \pi \cdot R_{average}}$$

$\neq$   
 B<sub>1</sub>      B<sub>2</sub>  
 $\equiv$

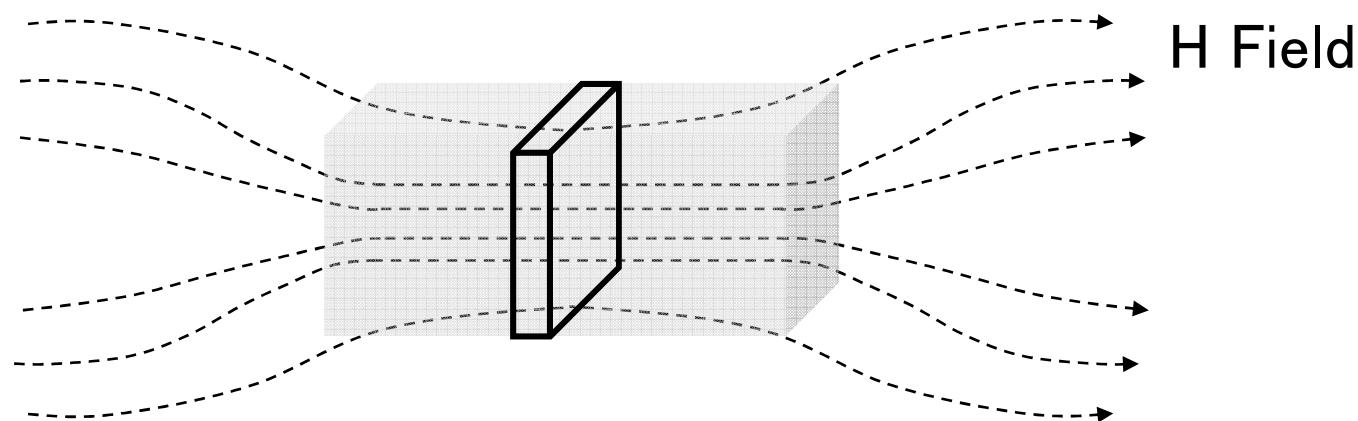
$$B = \mu_0 H$$

# Permeability



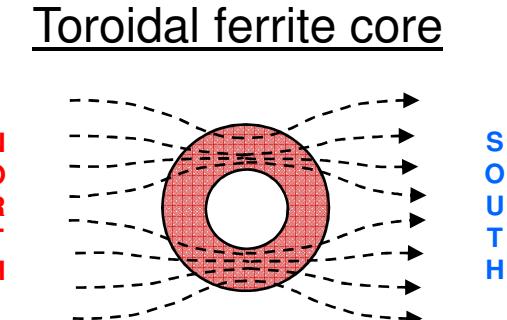
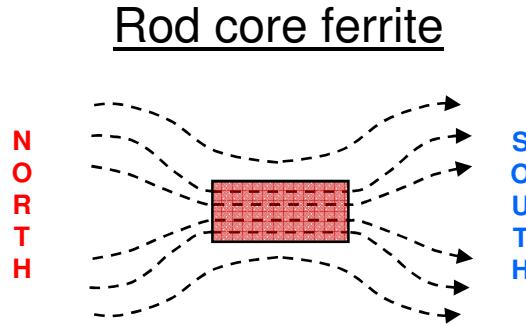
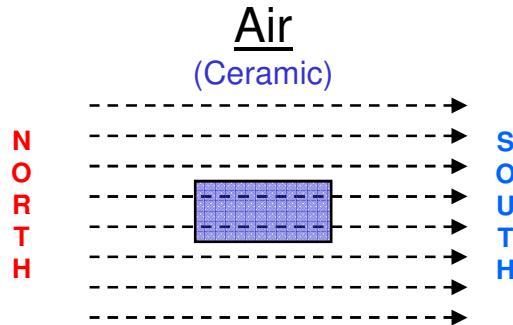
## Permeability

Describes the capacity of concentration of the magnetic flux in the material  
It is a factor of the energy needed to magnetize a material (i.e. magnetically align dipoles)



$$B = \mu_0 H$$

# Permeability



Induction in air:

$$B = \mu_0 \cdot \mu_r \cdot H$$

↑

Linear function because  $\mu_r = 1$

Induction in Ferrite:

$$B = \mu_0 \cdot \mu_r \cdot H$$

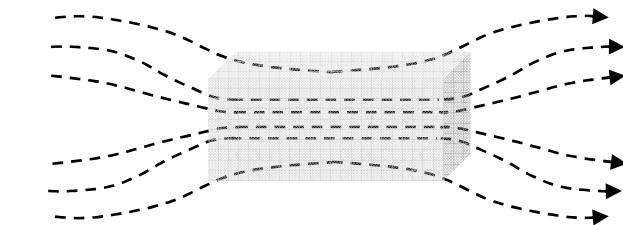
↑

Relative permeability is non-linear

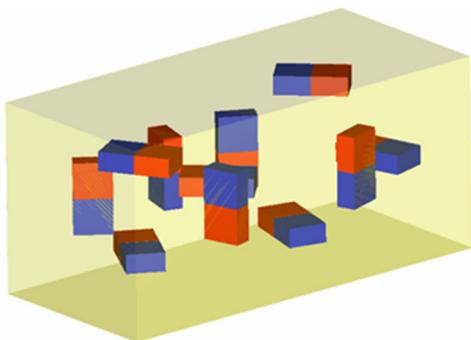
$$B = \mu_0 \cdot H$$

<b>Material</b>	<b>Frequency</b>	<b>Temperature</b>
<b>Current</b>	<b>Pressure</b>	

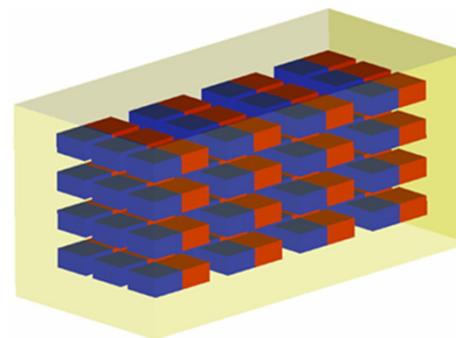
# Magnetic Materials



H Field



Non-magnetic  
Insulators

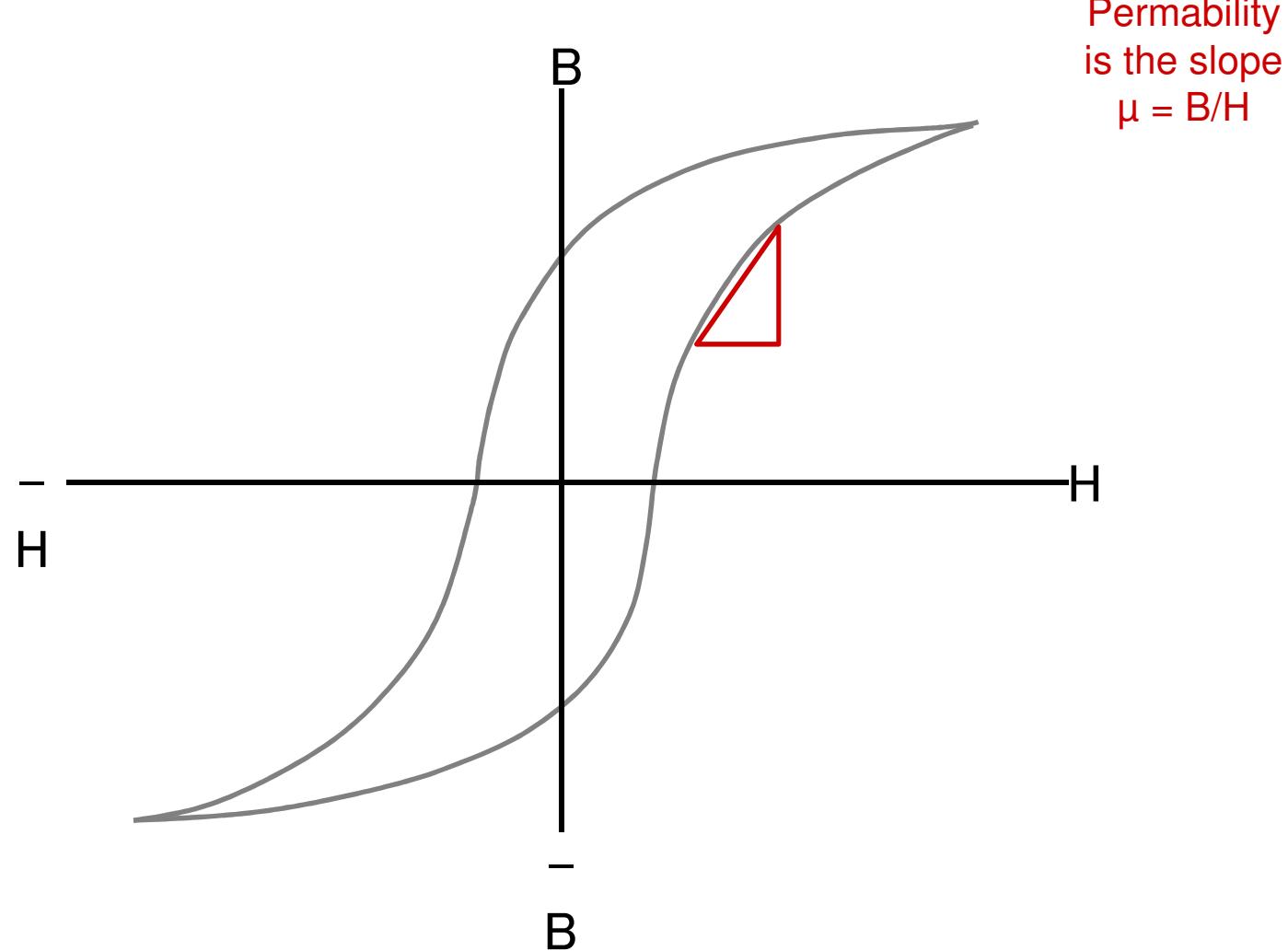


Soft-magnetic  
Ferrites



Hard-magnetic  
Fe Alloys

# Hysteresis Graph



## Soft Ferrites ( Have low losses at high frequencies)



- NiZn and MnZn ferrites can be easily magnetized and demagnetized **without dissipating much energy** (hysteresis loss)
- Hysteresis loss is a component of the total core loss.
- Another component of total core loss are Eddy currents which leak useful magnetizing energy
- MnZn has a higher permeability and saturation induction than NiZn
- NiZn has higher resistivity than MnZn. ( NiZn works well above 1 MHz)



# Hard Ferrites ( Have high losses in general)



- Hard ferrites cannot be easily magnetized and demagnetized. They need a substantial amount of energy change the direction of the magnetization ( high hysteresis loss)
- Iron-based ferrites have the general formula  $MO\text{-Fe}_2\text{O}_3$  where M is a divalent ion such as Fe, Ni, Cu, Mg, Mn, Co, Zn or Li.
- They are great conductors of magnetic flux well and have a high magnetic permeability
- They store ( handle) stronger magnetic field than soft ferrites
- Relatively, they have low resistivity

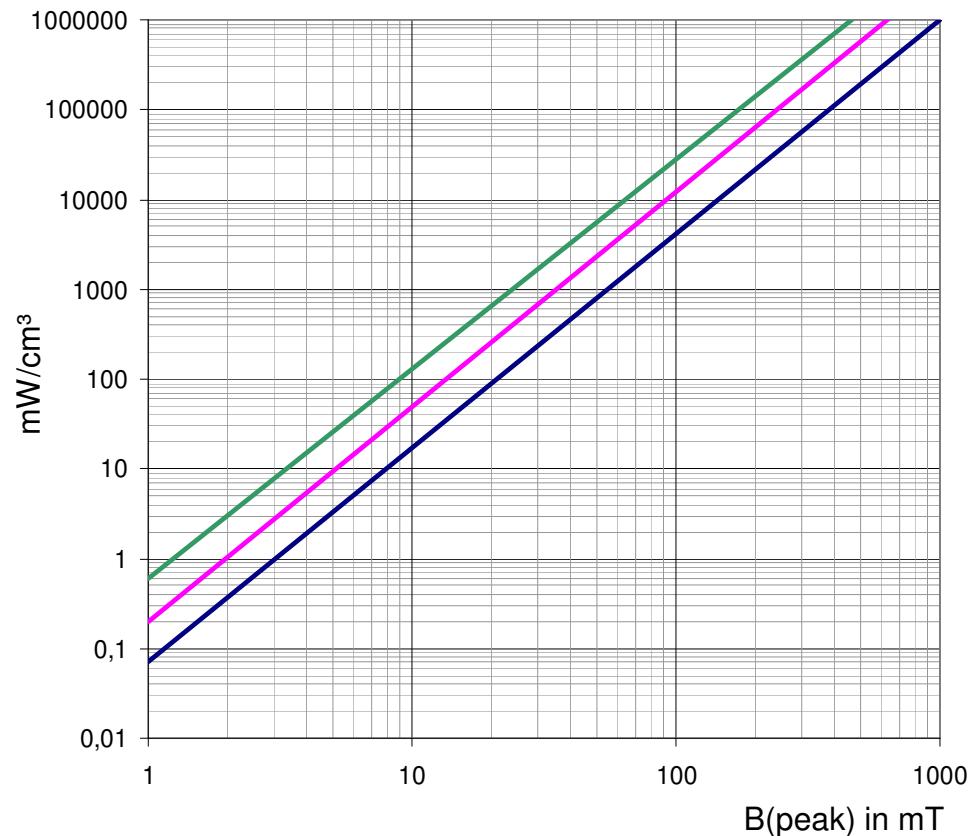


# Core losses

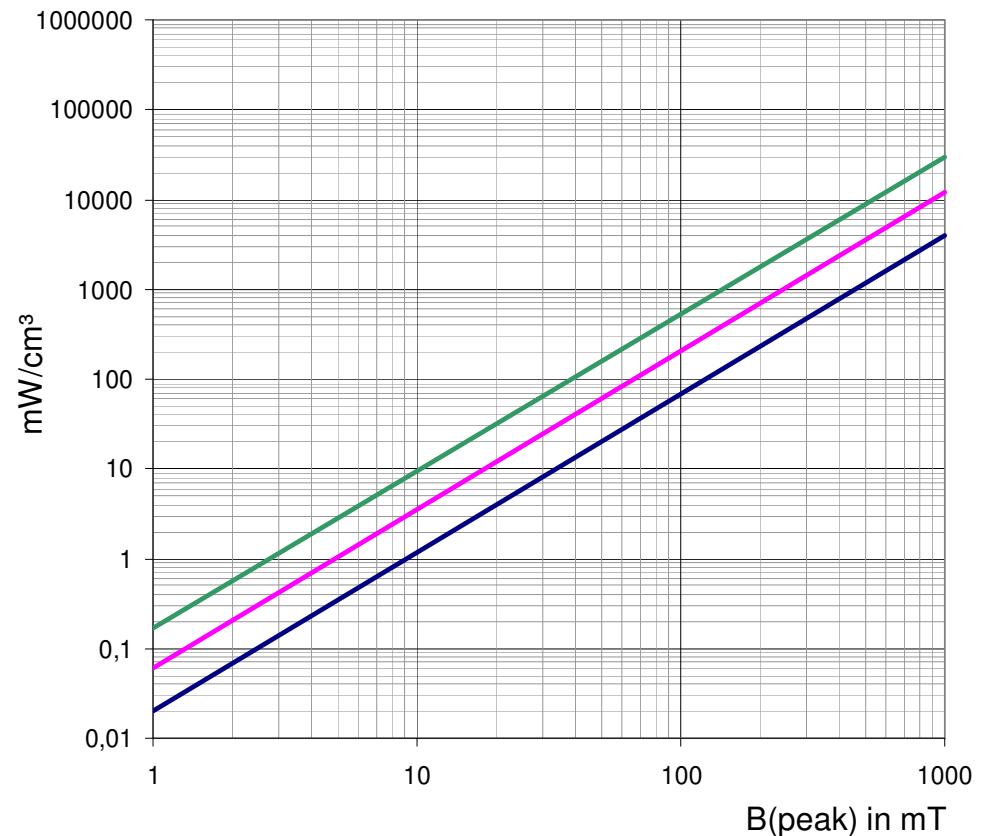
## Comparing materials

500KHz  
250KHz  
100KHz

Iron powder



NiZn

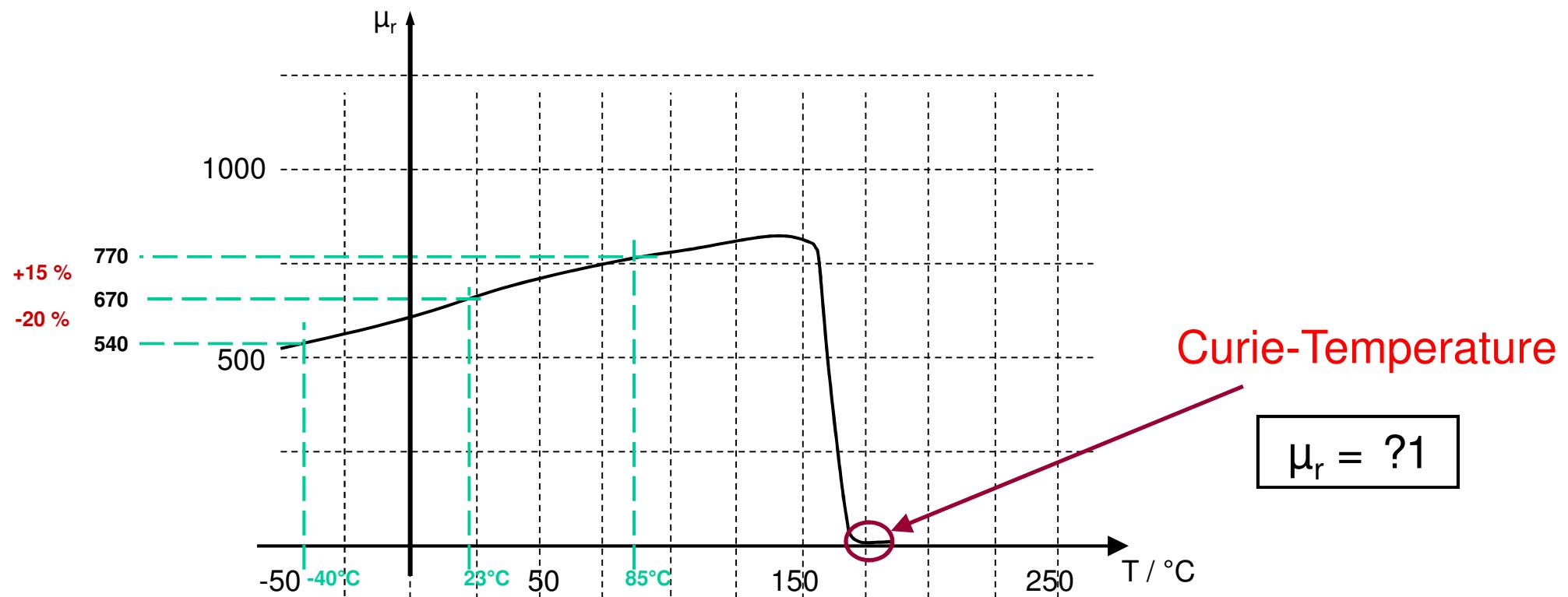


# CORE GAP, INDUCTOR IMPEDANCE, TEMPERATURE PERFORMANCE AND SHIELDING

## Permeability's dependence on temperature

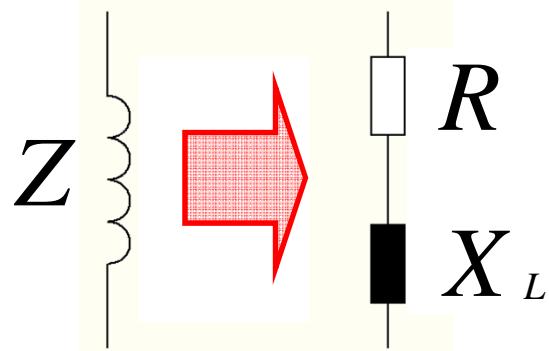
$T \uparrow \rightsquigarrow$  thermal motion  $\uparrow \rightsquigarrow$  degree of order  $\downarrow$

## Alignment of elementary magnets



# Complex Permeability

## Equivalent Circuit diagram



$$Z = \sqrt{R^2 + X_L^2}$$

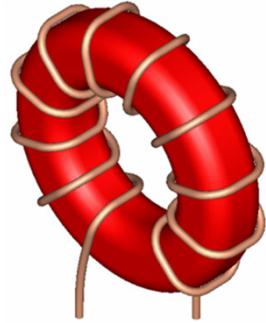
X<sub>L</sub> is called the reactance ( a measure of storing ability)

R is called the resistance ( a measure of dissipating ability)

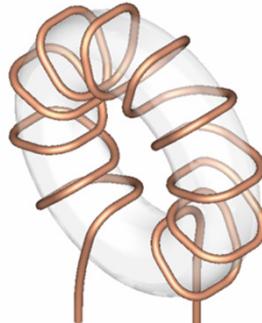
X<sub>L</sub> is purely inductive and hence stores and releases energy without any loss

R is purely lossy and hence it dissipates energy

# Permeability – complex permeability



=



•



Impedance of winding on  
with core material

=

Impedance of winding  
w/o core material

•

core material

$\underline{Z}$

=

$j \omega L_0$

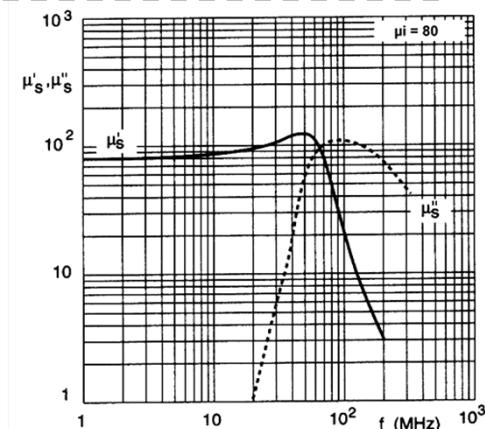
$(\mu^{\parallel} - j\mu^{\perp})$

$R$

=

$\omega L_0$

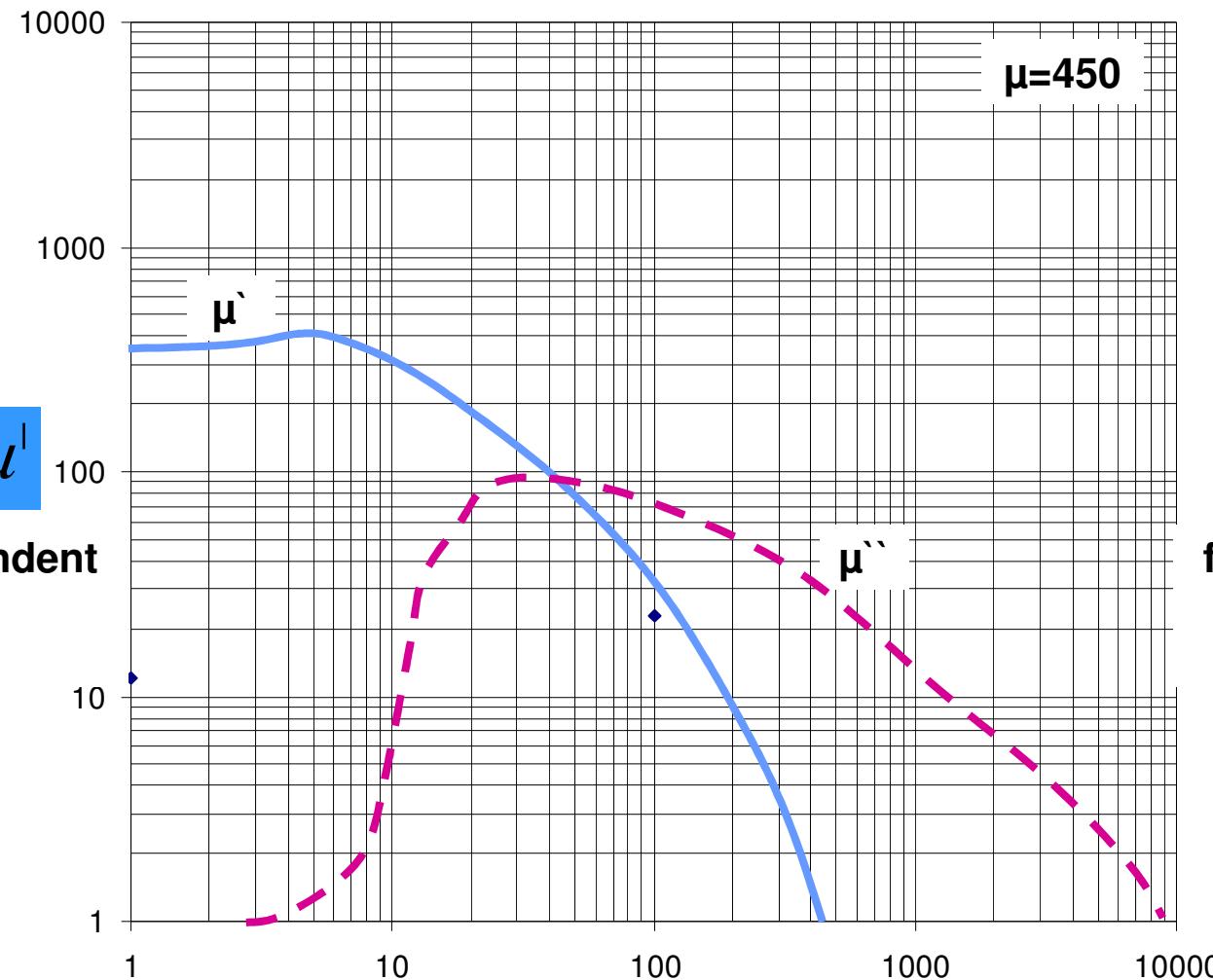
•



# Permeability – complex permeability



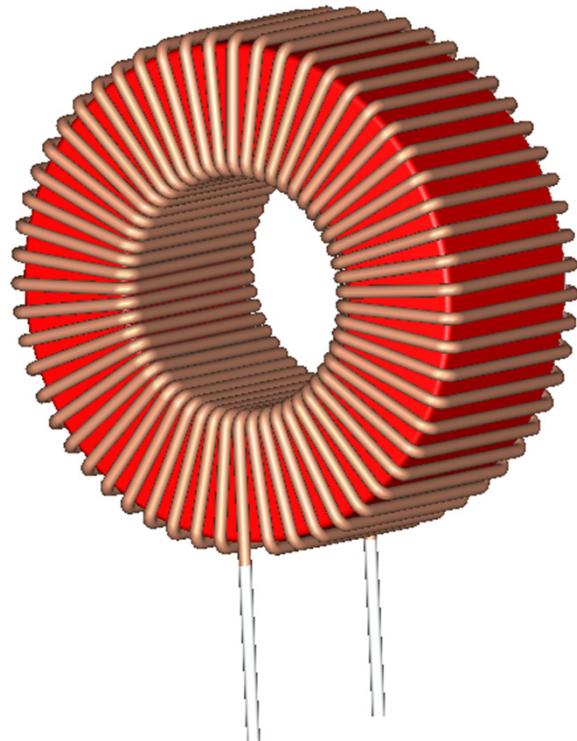
$$Z = j\omega L_0 (\mu^{\parallel} - j\mu^{\perp}) = R + jX$$



# Inductance and Core Material Geometry



## Toroidal Inductor



$$L = \frac{(\mu_0 * \mu_r * A_{\text{eff}} * N^2)}{l_{\text{eff}}}$$

$$\mu_0 = 4 \cdot \pi \cdot 10^{-7}$$

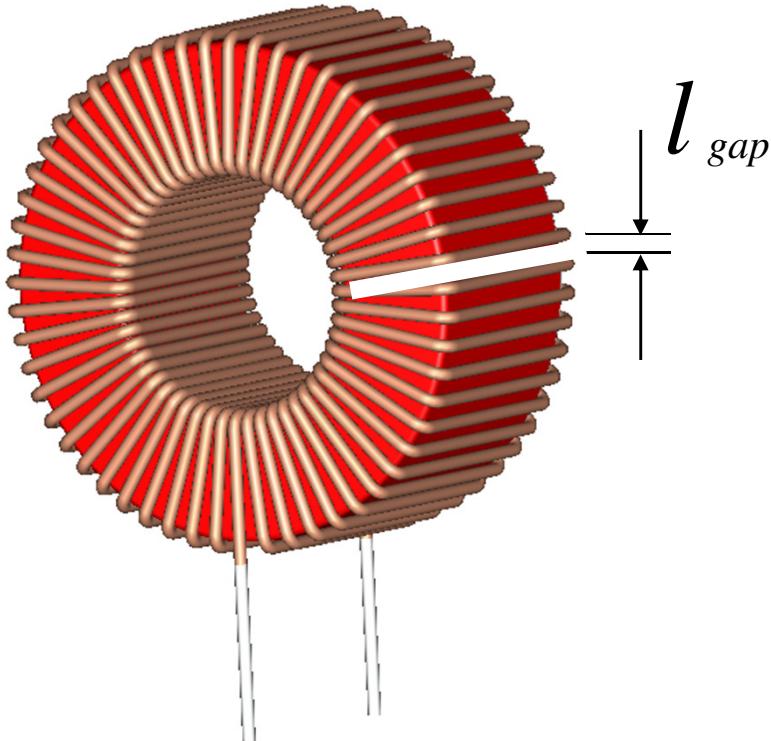
$\mu_r$  = relative permeability  
 $N$  = No. of turns  
 $A_{\text{eff}}$  = effective magnetic area  
 $l_{\text{eff}}$  = effective magnetic length

# Inductance and Core Material Geometry



## Toroidal Inductor with a gap

Adding an air gap will increase the effective magnetic length

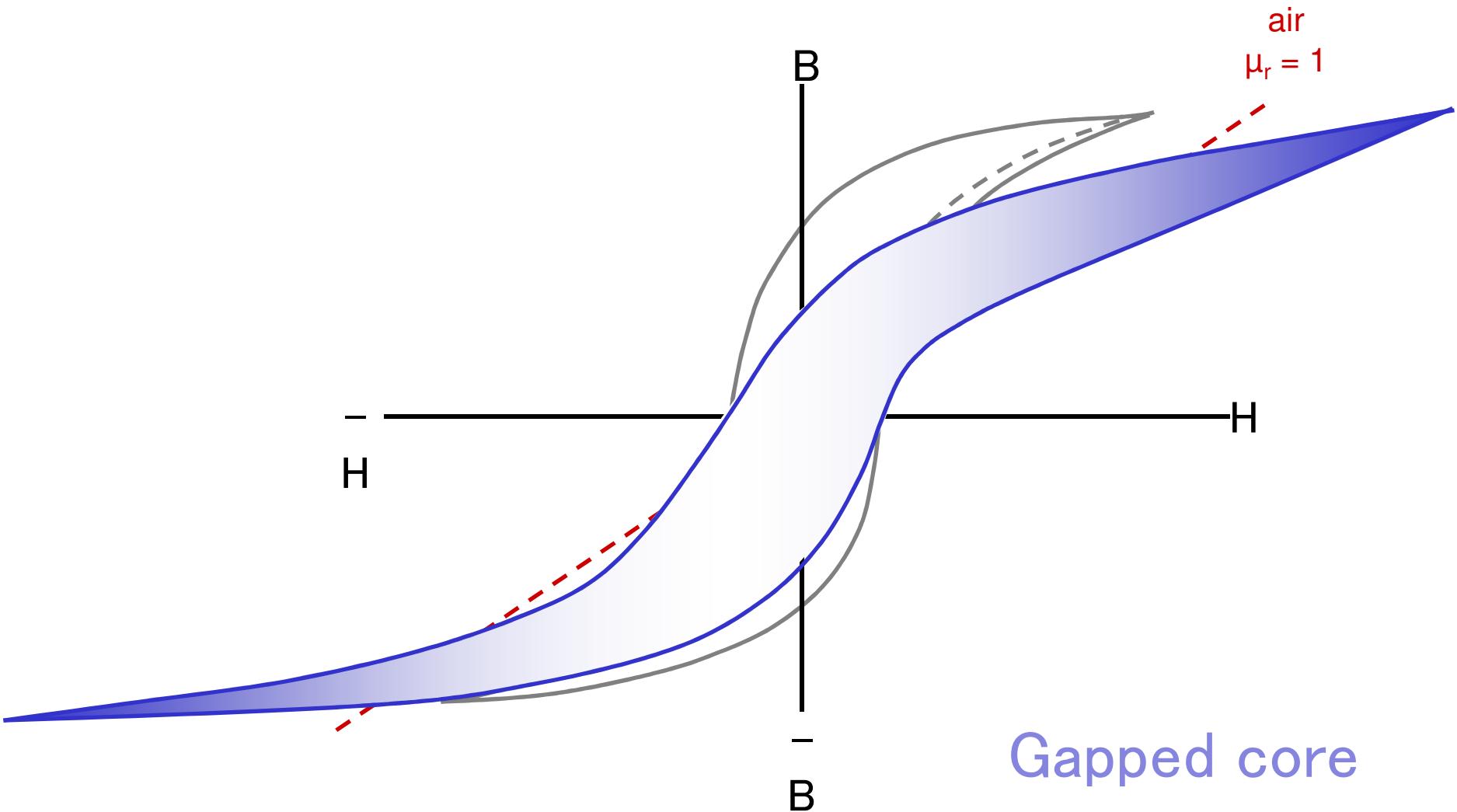


$$L = \frac{(\mu_0 * \mu_r * A_{eff} * N^2)}{l_{eff}}$$

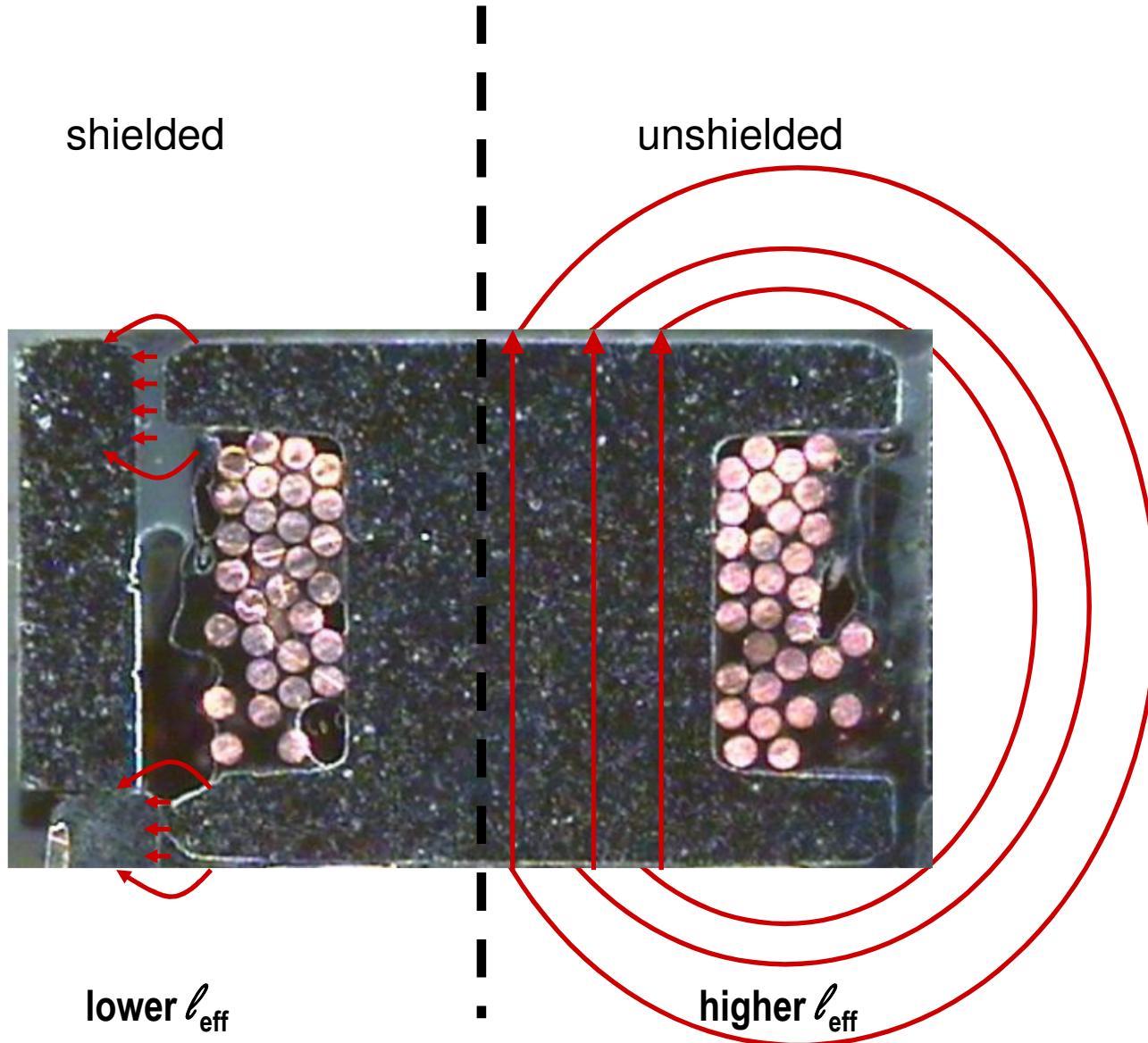
The effective magnetic length  $l_{eff}$  is shown as a dashed line that follows the path of the magnetic field, which is longer than the physical length of the core due to the air gap.

$$l_{eff} \sim l_{eff} + l_{gap} \mu_r$$

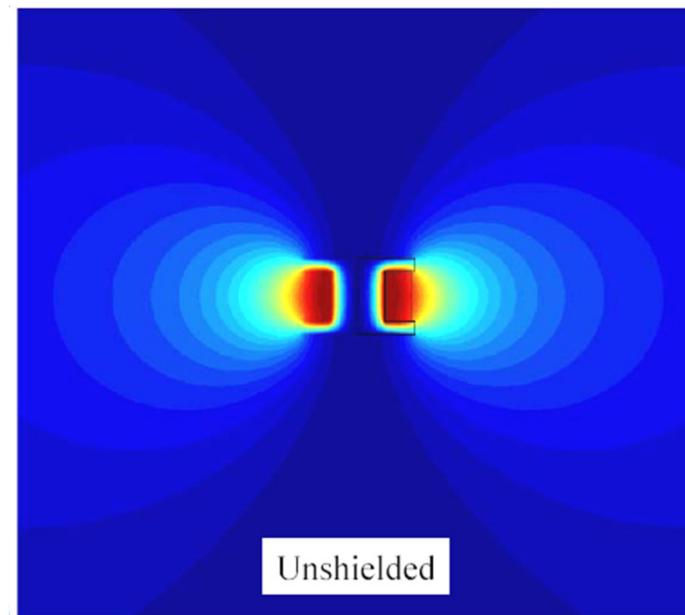
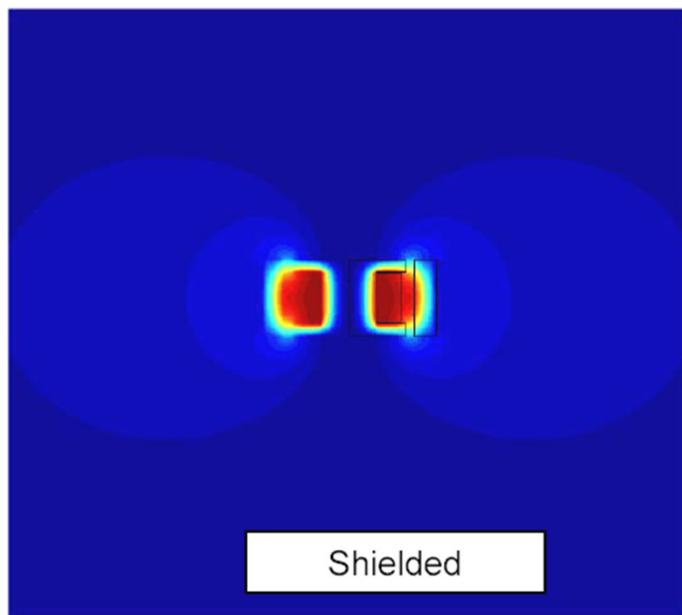
## Un-gapped Cores Vs. Gapped cores



# Shielded Vs. Unshielded



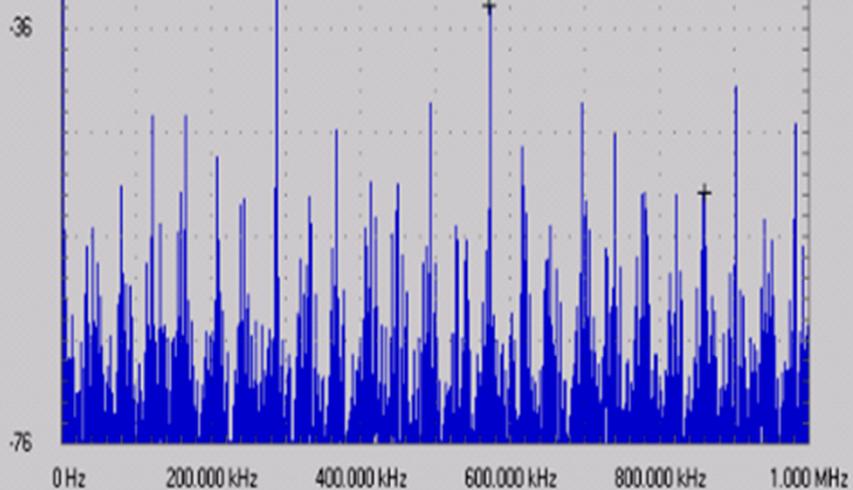
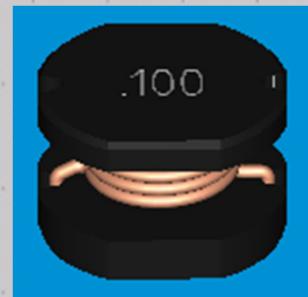
# Stray field - Shielded vs. Unshielded



# stray field - shielded vs. unshielded



unshielded



Channel : 2

Window Type : Blackman

Window Size : 8192

Cursor Information : Cursor off

Harmonic Information :

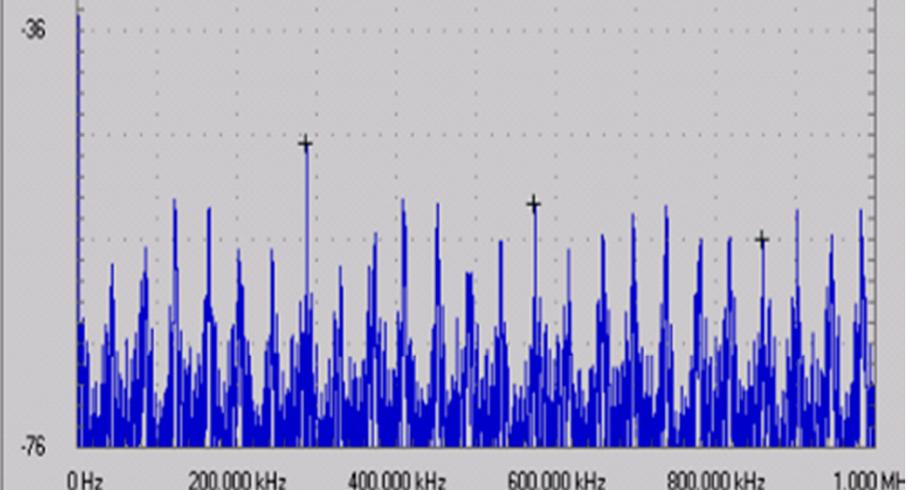
1st: 286.750 kHz -26.856 dB

2nd: 573.250 kHz -33.947 dB

3rd: 860.000 kHz -51.847 dB

4th: --

shielded



Channel : 2

Window Type : Blackman

Window Size : 8192

Cursor Information : Cursor off

Harmonic Information :

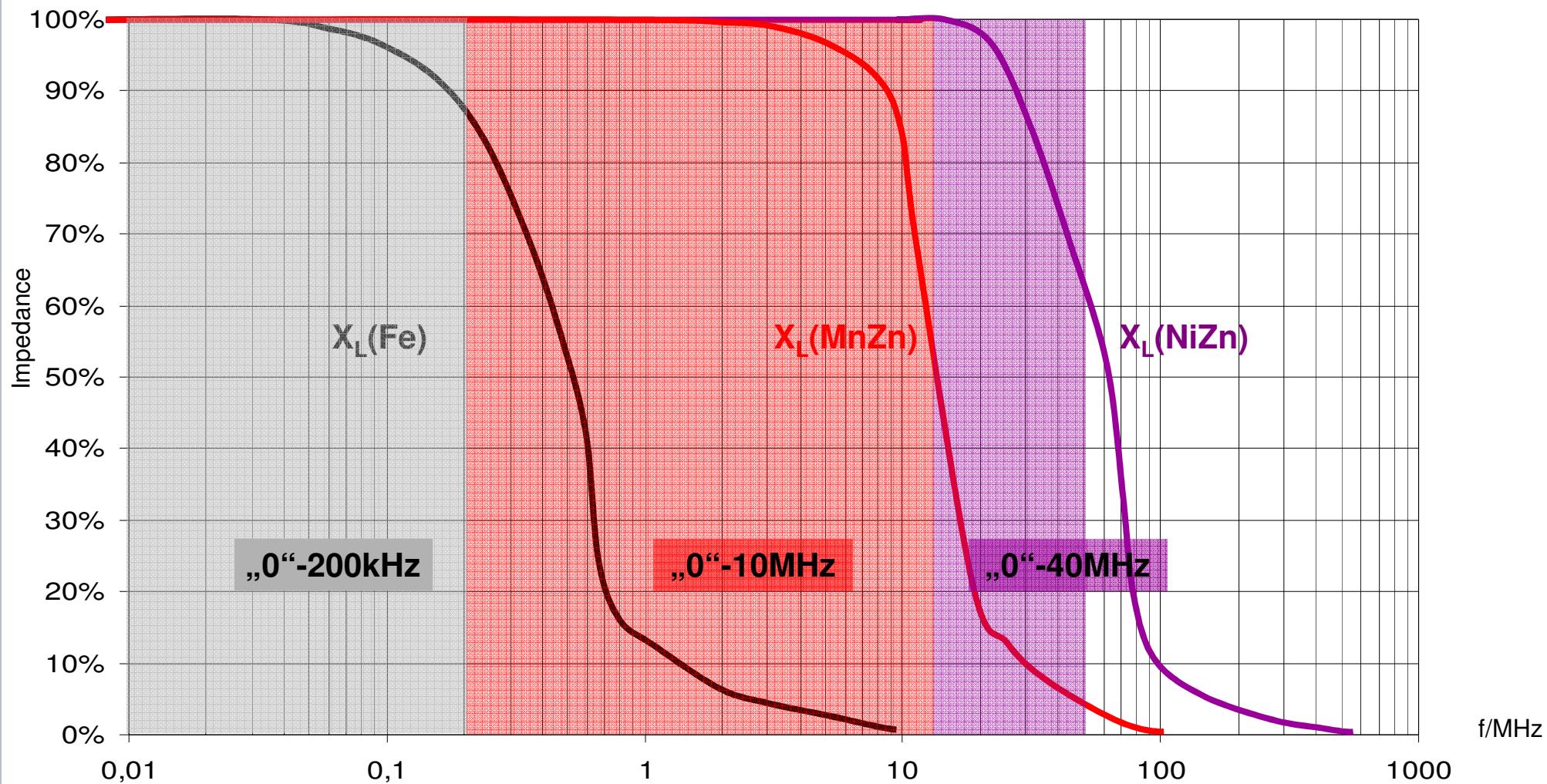
1st: 286.500 kHz -46.977 dB

2nd: 573.000 kHz -52.705 dB

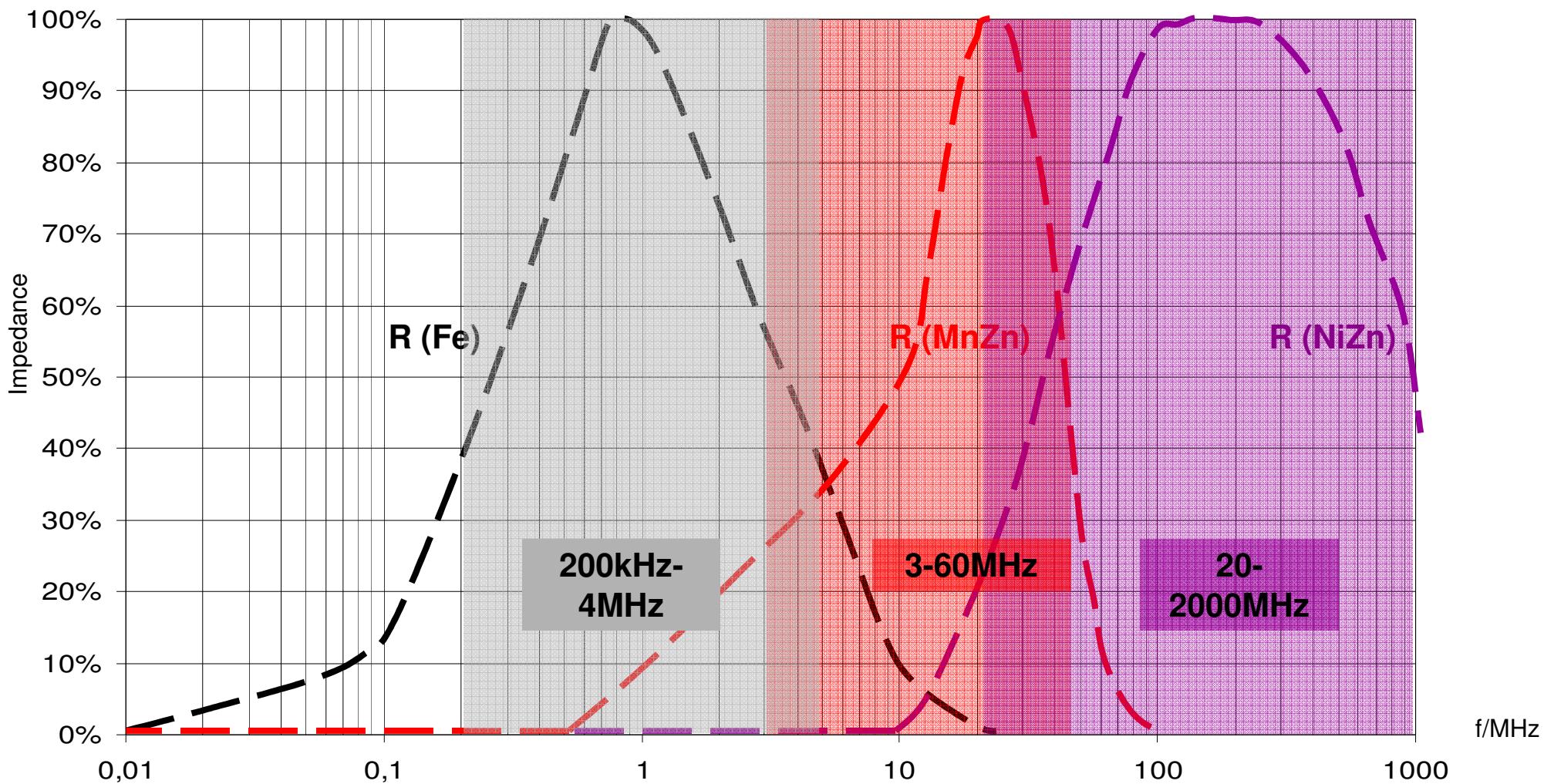
3rd: 859.750 kHz -56.160 dB

4th: --

## Applications ( Inductors for Storage)



## Applications ( Inductors for Filtering)

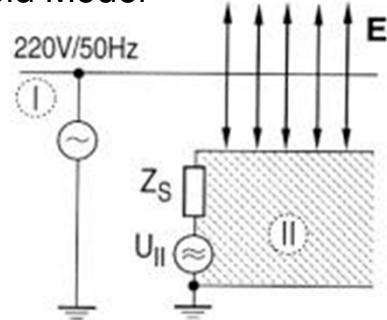


# Types of Noise Coupling Mechanisms



## Capacitive Coupling

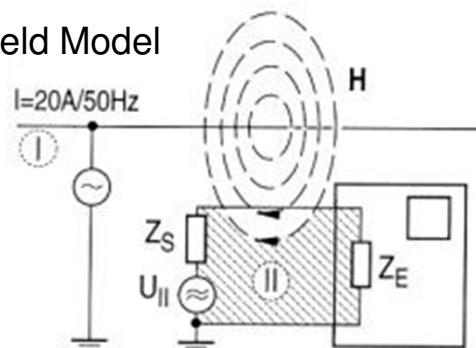
Field Model



Evident in cases where components or traces are at distance of  $1/_{10} \lambda$  (wavelength) of the **signal frequency**

## Inductive Coupling

Field Model



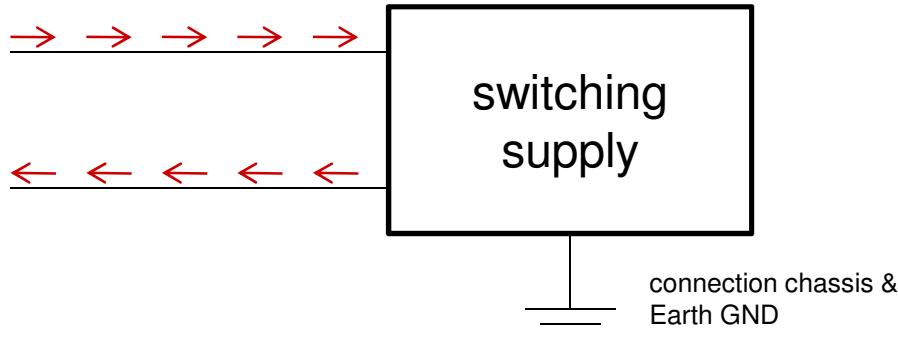
Evident in cases where components or traces are at a distance of  $1/4 \lambda$  (wavelength) of the **noise signal**

## COMMON MODE CHOKES

# Types of Noise Signals

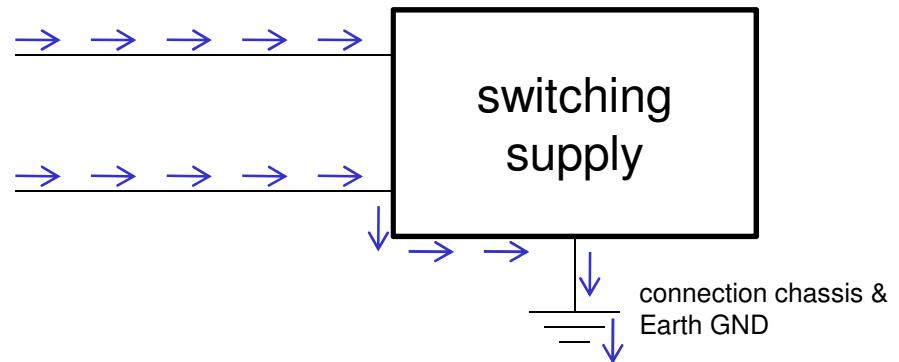
## Differential-Mode signal

- Noise flows into one line and exits through another
- Independent from GND



## Common-Mode signal

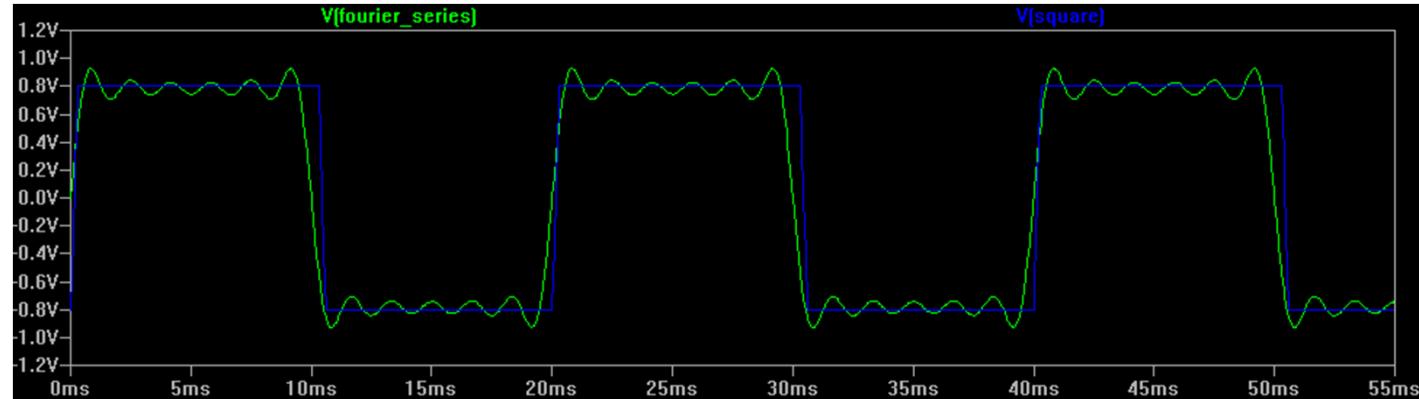
- Noise flows along both lines in the same direction
- returns by some parasitics path through system GND



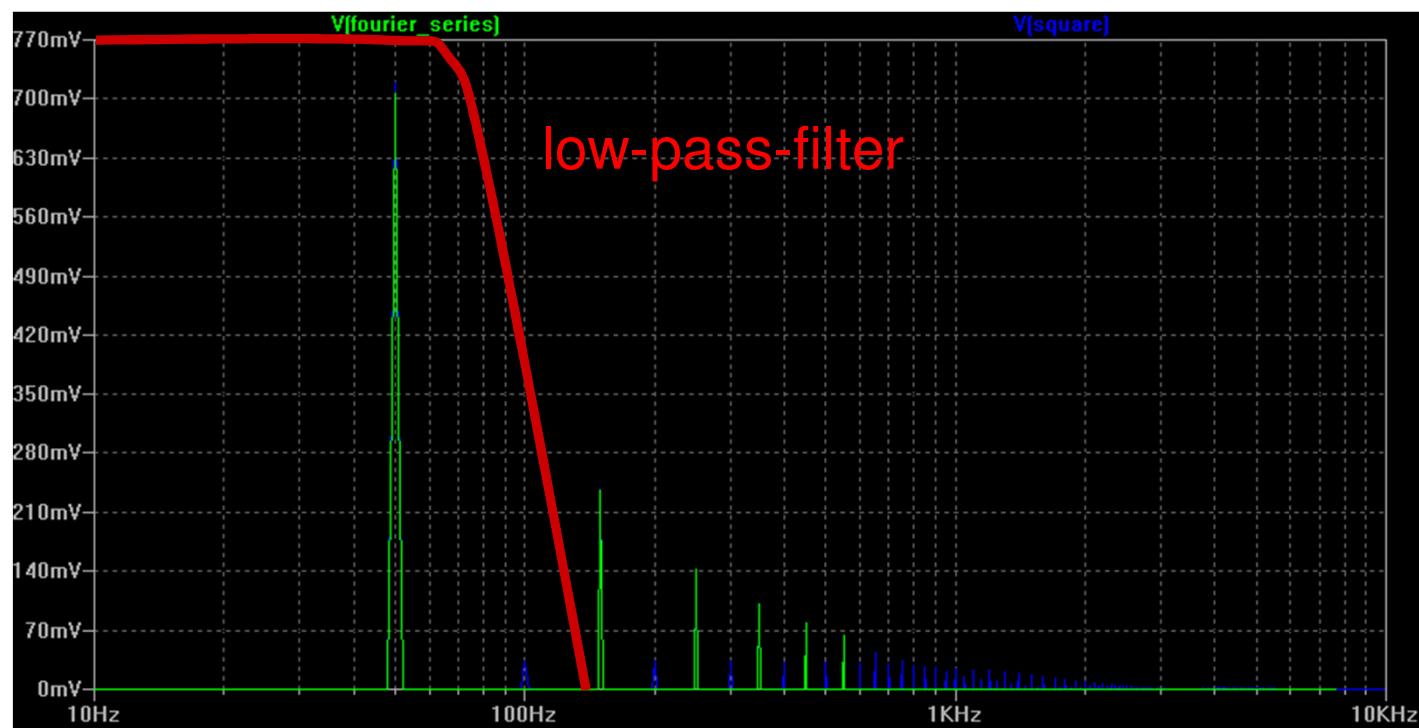
# Common mode choke - advantages



time based



frequency based

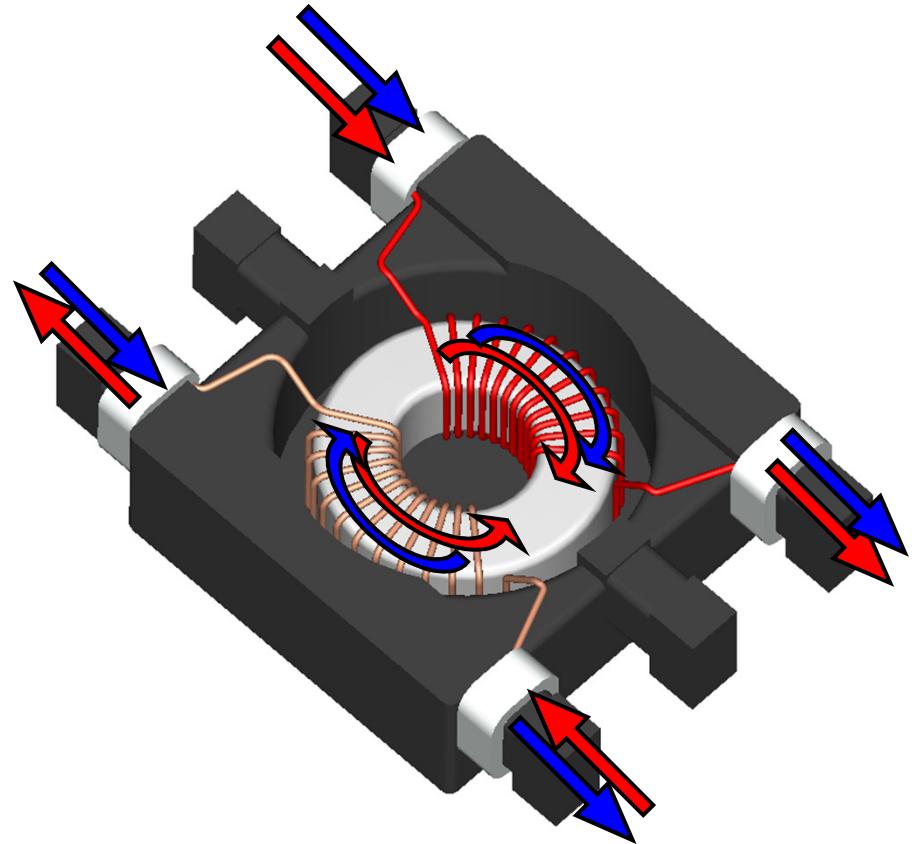


# Common Mode Filter



## Reduction of noise

- From device to environment
- From environment to device



## Conclusion:

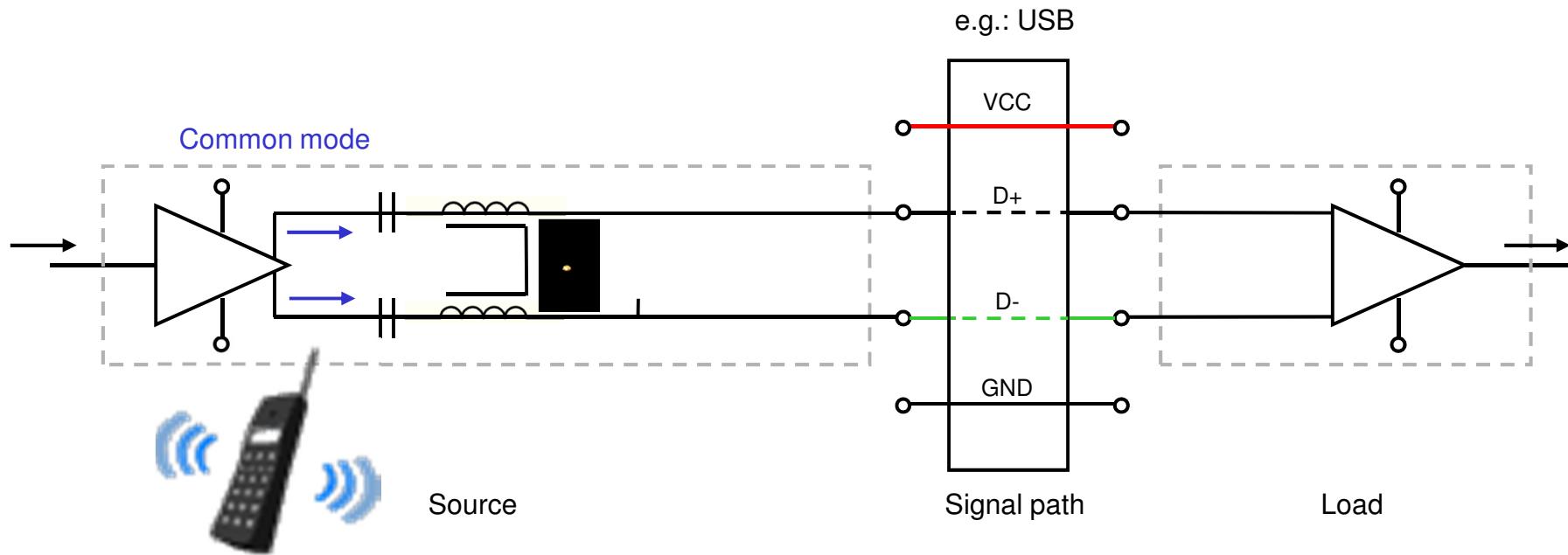
- “Almost” no influencing of the signal  
→ **Differential mode**
- High attenuation of noise

→ Common mode

# Common Mode Filter – Signal theories



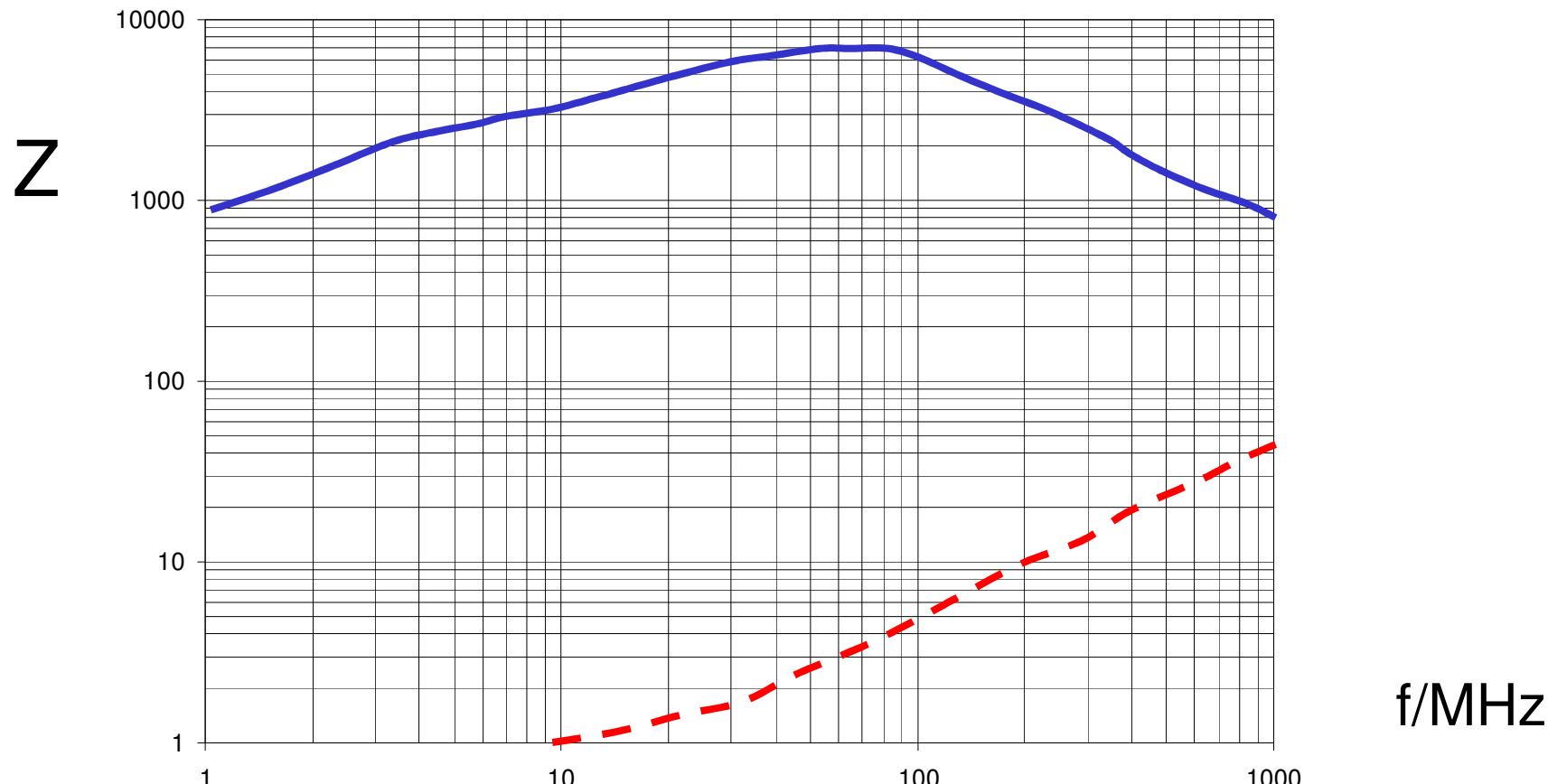
## Filtering



## Common Mode Filter



- The Common Mode Impedance attenuates just the noise



- The Differential mode-impedance will also attenuate the signal

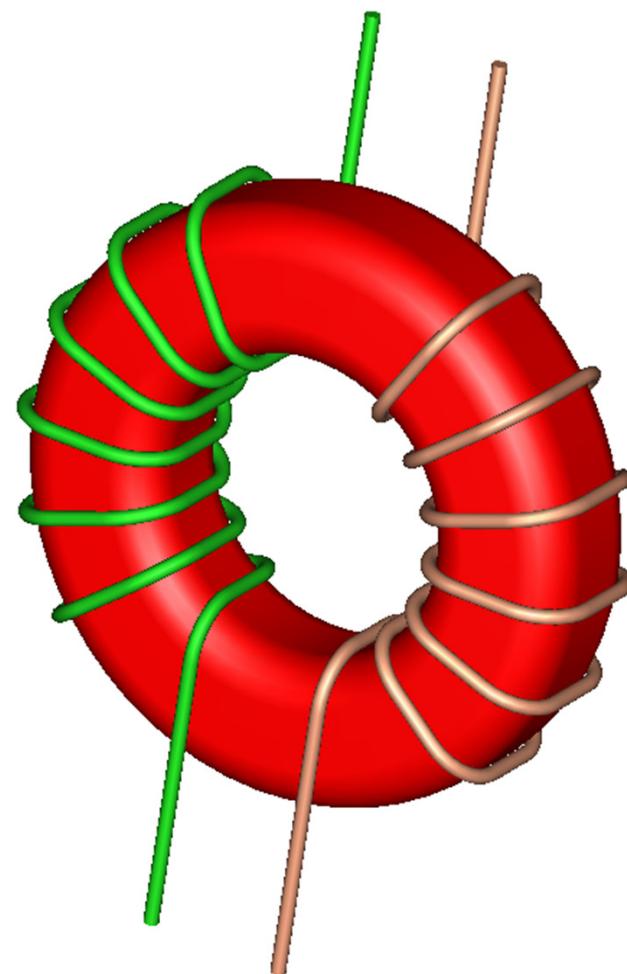
## Common Mode Choke Construction



## bifilar



## sectional



# Common mode choke - construction

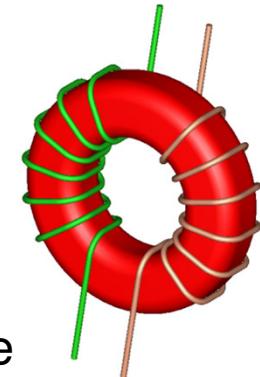


## Bifilar

- Less differential impedance
- High capacitive coupling
- Less leakage inductance

## Sectional

- Low capacitive coupling
- High leakage inductance
- High differential impedance



- Data lines  
→ USB, Fire-wire, CAN, etc.
- Power supply
- Measuring lines
- Sensor lines

- Power supply input /output filter  
→ CMC for mains power
- High voltage application
- Measuring lines
- Switching power supply decoupling



• WE-CNSW



• WE-SLM



• WE-LF  
• WE-SLx-Series

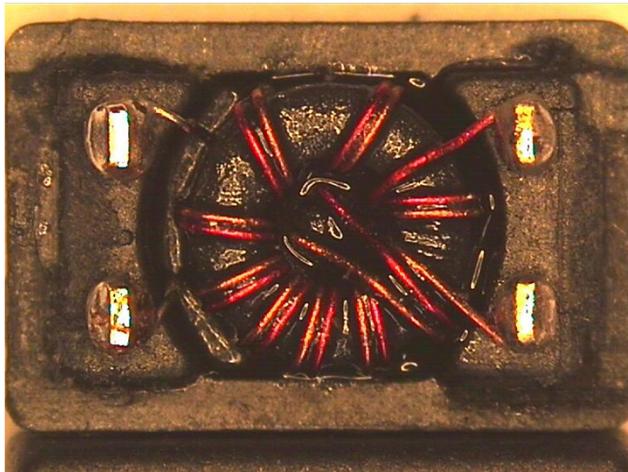


• WE-CMB  
• WE-VB / VB2

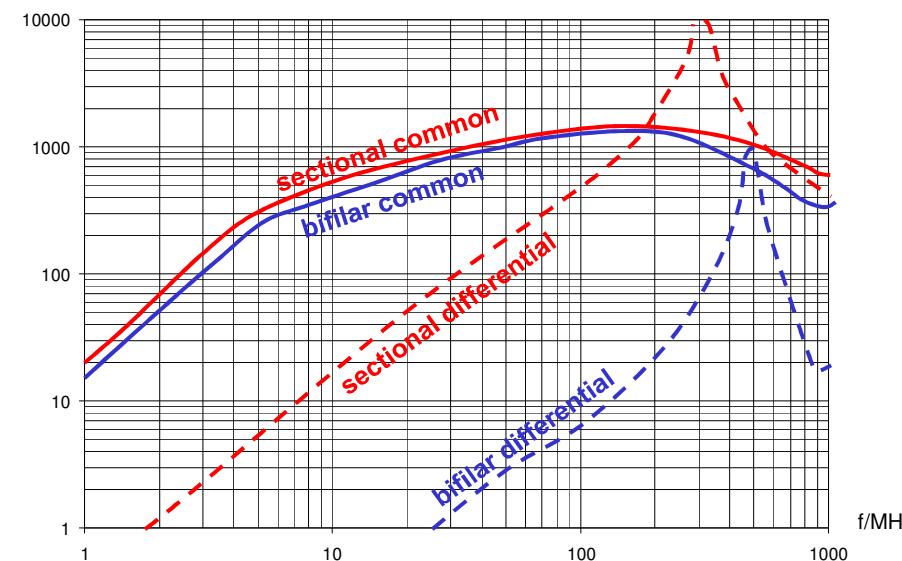
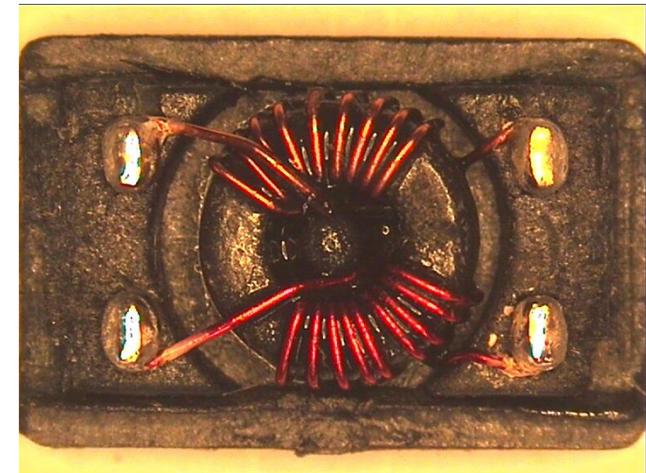
# Common mode choke - construction



## bifilar winding



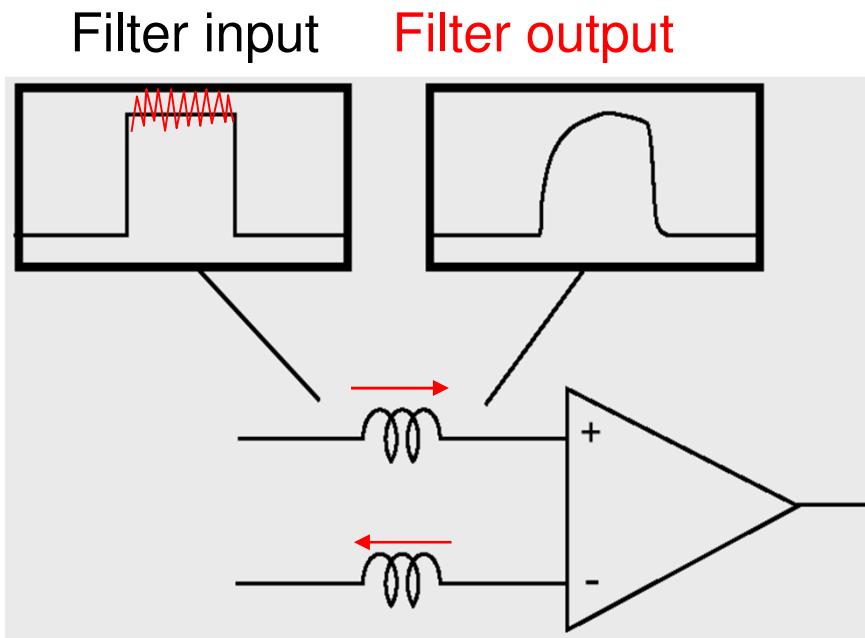
## sectional winding



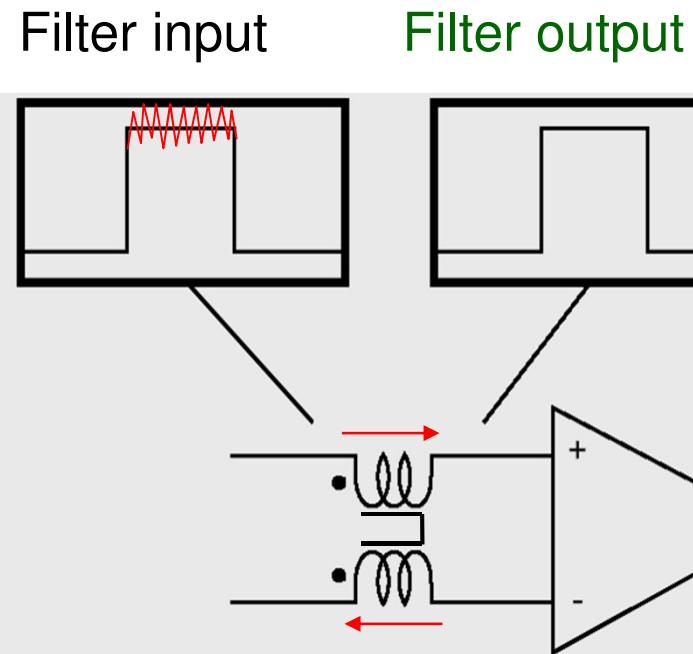
# Common mode choke - advantages



## Filter with two inductors

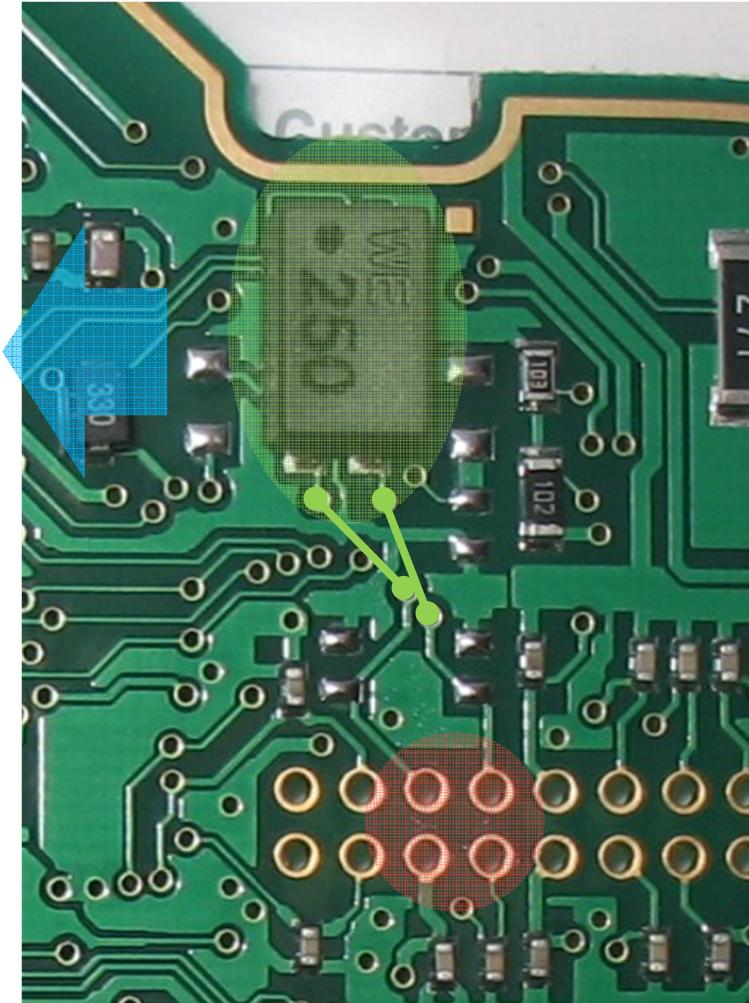


## Filter with CMC



- no signal distortion – means signal will be not affected

# Common mode choke - advantages



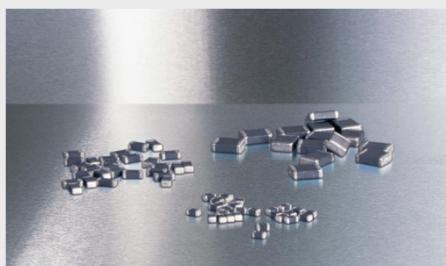
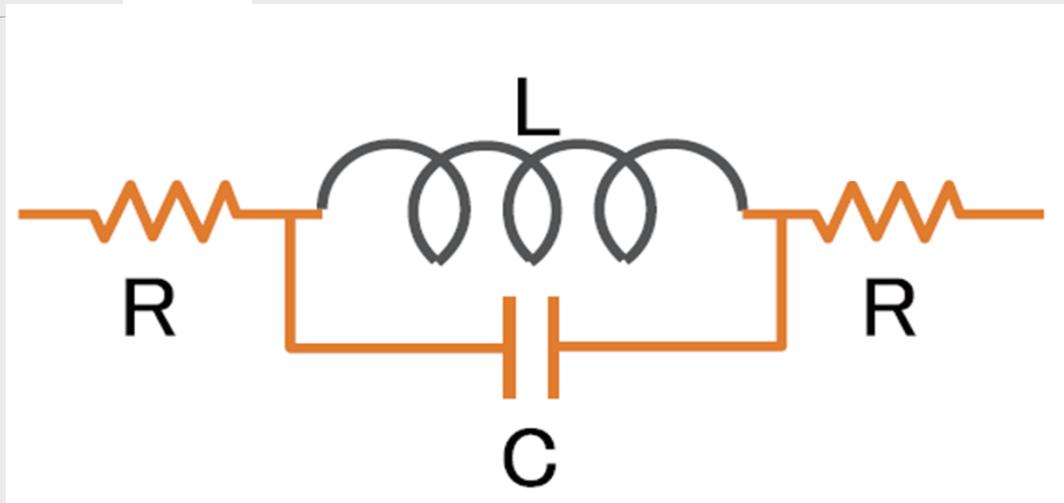
Digital logic device

Common Mode Choke

Interconnection → Cable/Connector

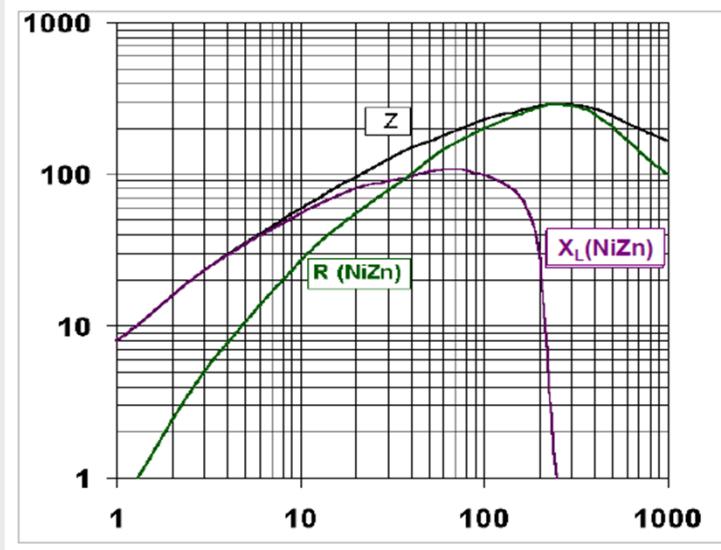
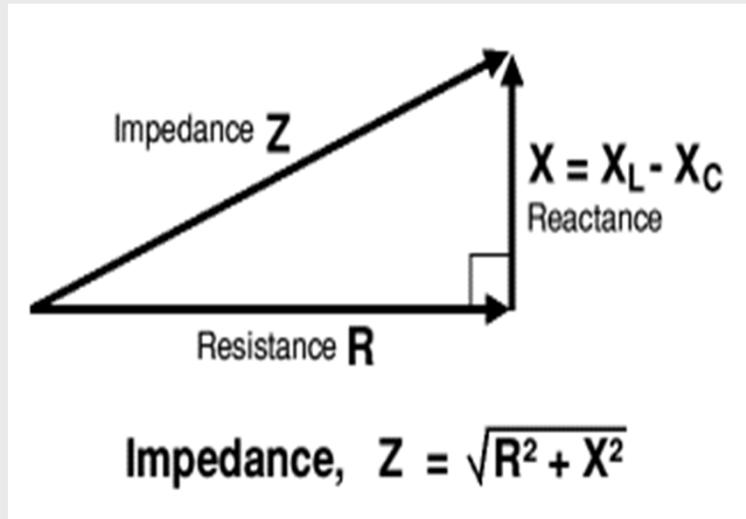
## CHIP BEAD FERRITES

# Chip Bead Ferrites



- **Inductor**
  - Used to filter unwanted noise
  - Supply voltage lines, ground planes, and data signals
- **Frequency Dependant**
  - Also known as a frequency dependent resistor

# Chip Bead Ferrites Impedance

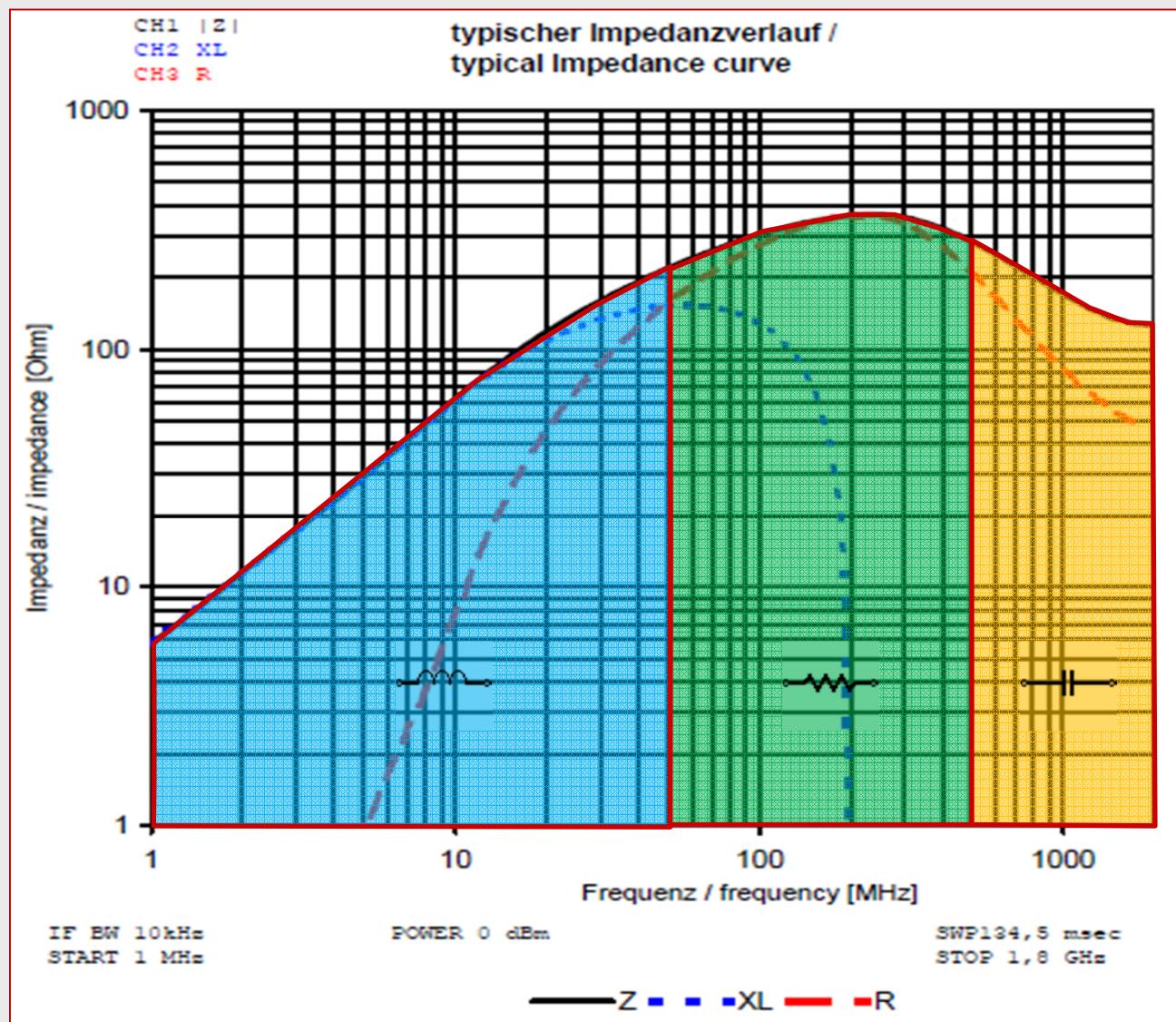


- AC resistance equivalent
  - Impedance can be split into two parts
    - Resistance R (Constant regardless of frequency)
    - Reactance X (Varies with frequency)

# Chip Bead Ferrites

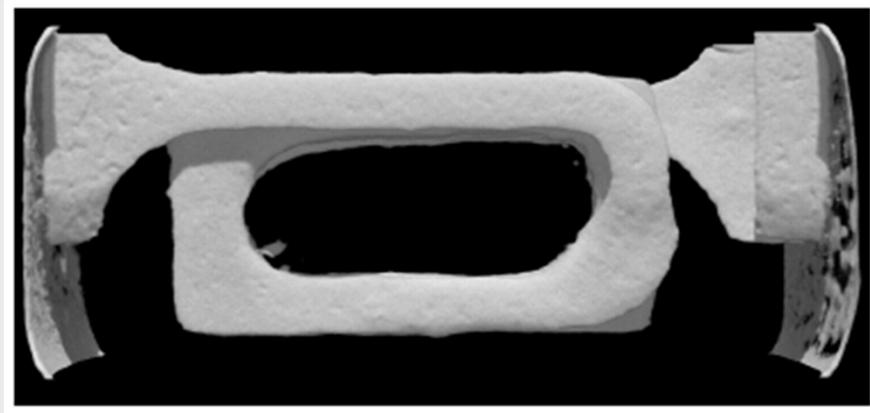
- **Resistive**
  - Impedes the noise energy and absorbs it
- **Inductive**
  - Energy storage
- **Capacitive**
  - Passes AC, blocks DC

SRF **XL=XC**

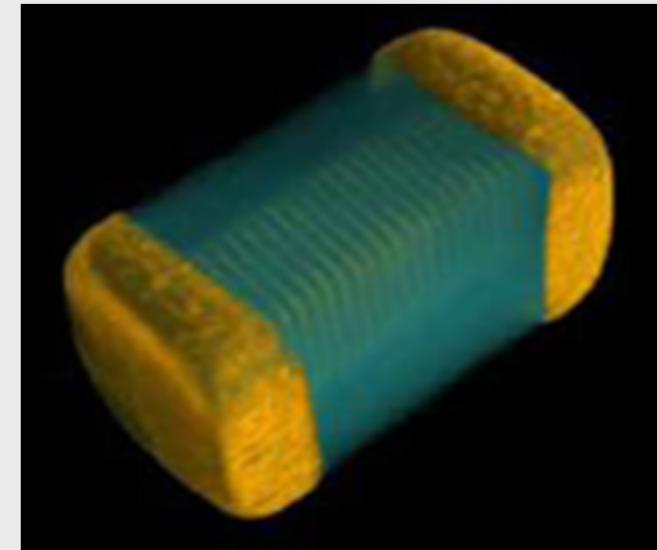
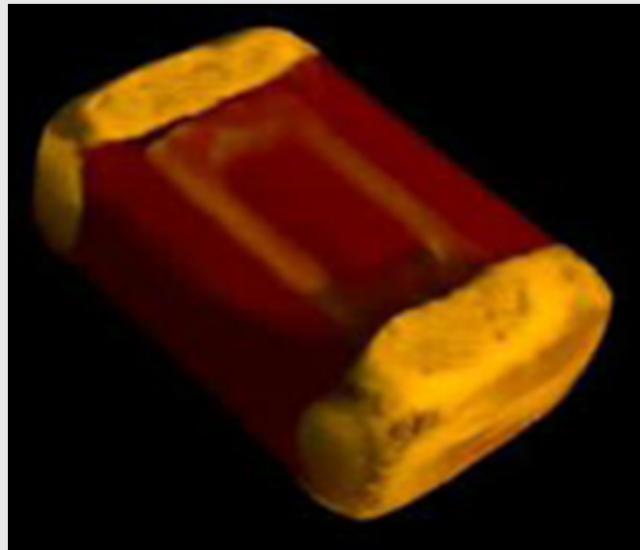
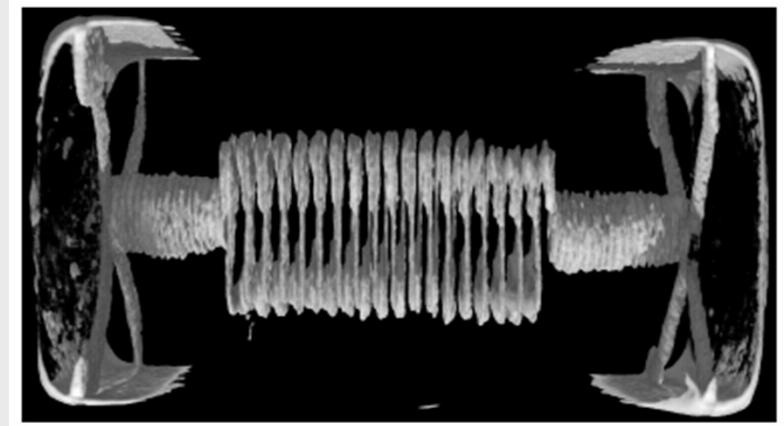


# Chip Bead Ferrites

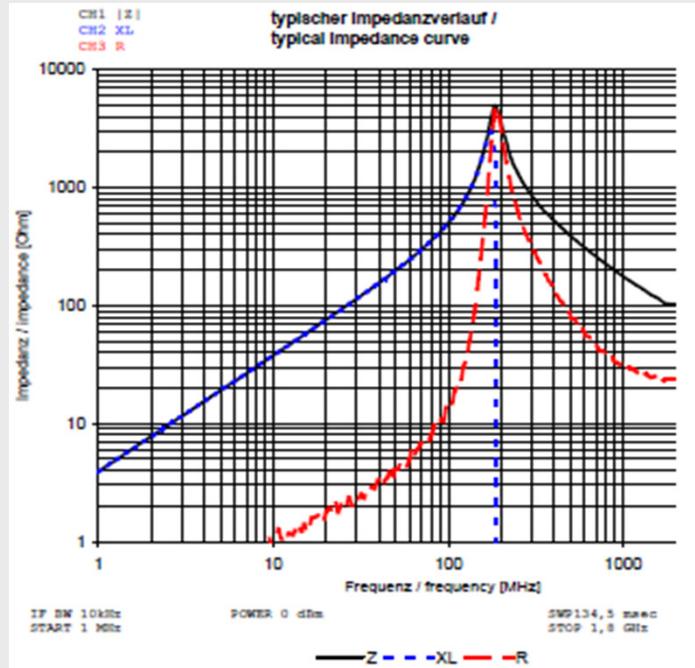
**High Current**



**High Frequency**

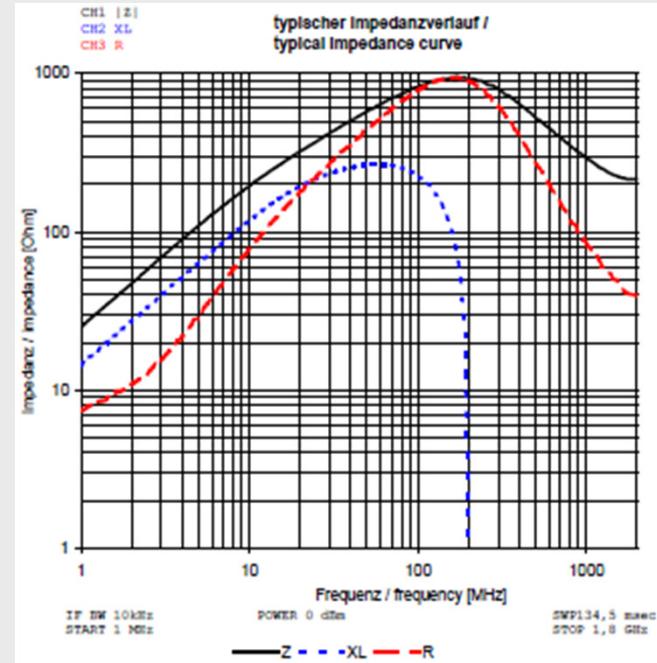


# Chip Bead Ferrites



## High speed

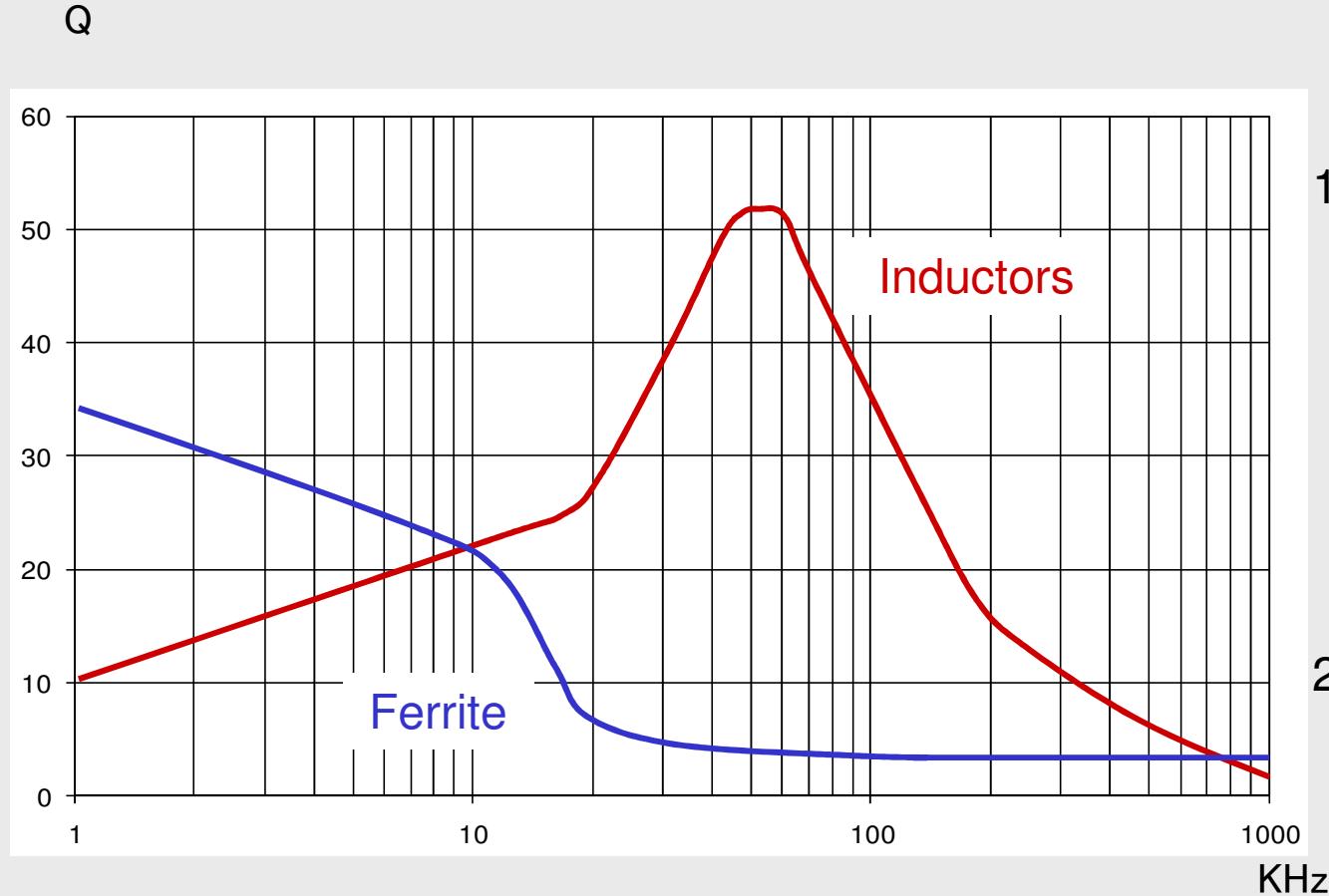
- Lower impedance in lower frequency
  - Low attenuation for fast signals
- Applications: USB 2.0, IEEE 1394, LVDS



## Wide band

- High impedance in low frequency
- Wide band through the entire spectrum
  - Applications: DC/DC converters

# Q factor Application

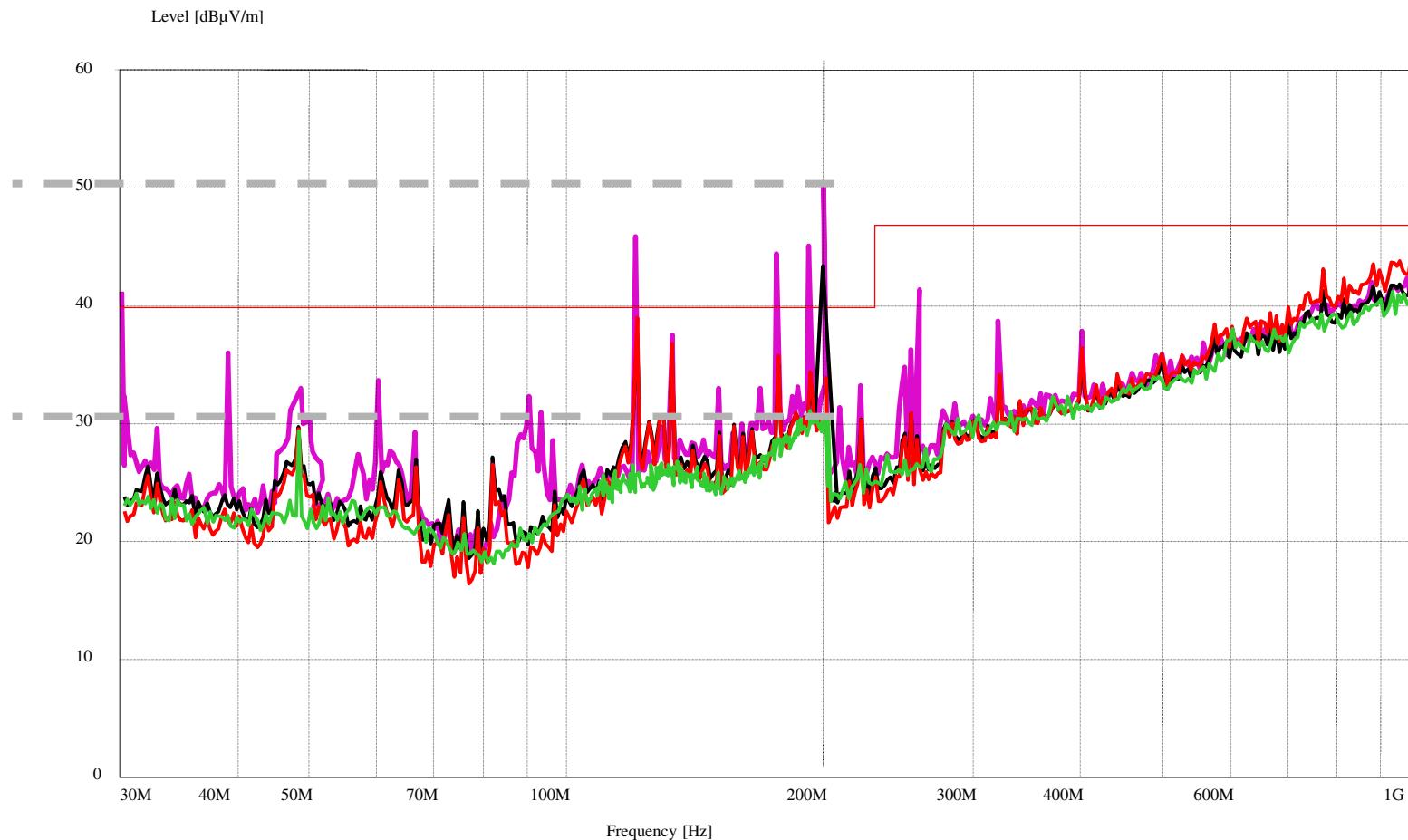


1. Application: **Storage inductor**
  1. Request: Lowest possible core losses at switching frequency (HIGH Q)
2. Application: Absorber / Filter
  1. Request: Highest possible core losses at application frequency (LOW Q)

## Insertion loss - example

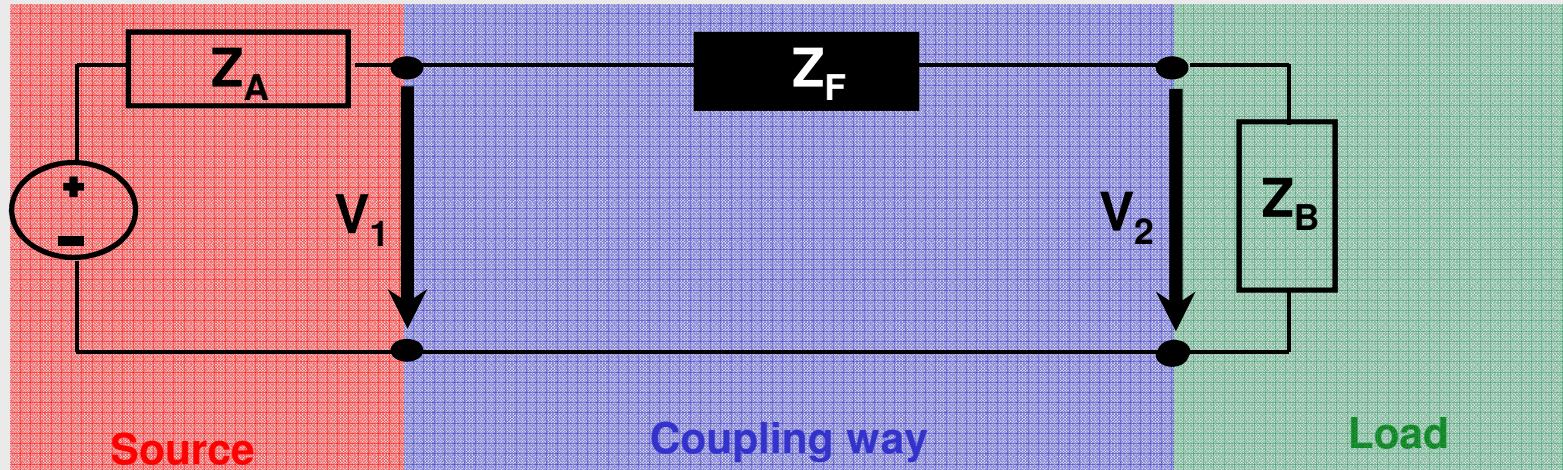


- check the results
- measuring the emission and compare the attenuation



# Insertion loss calculation

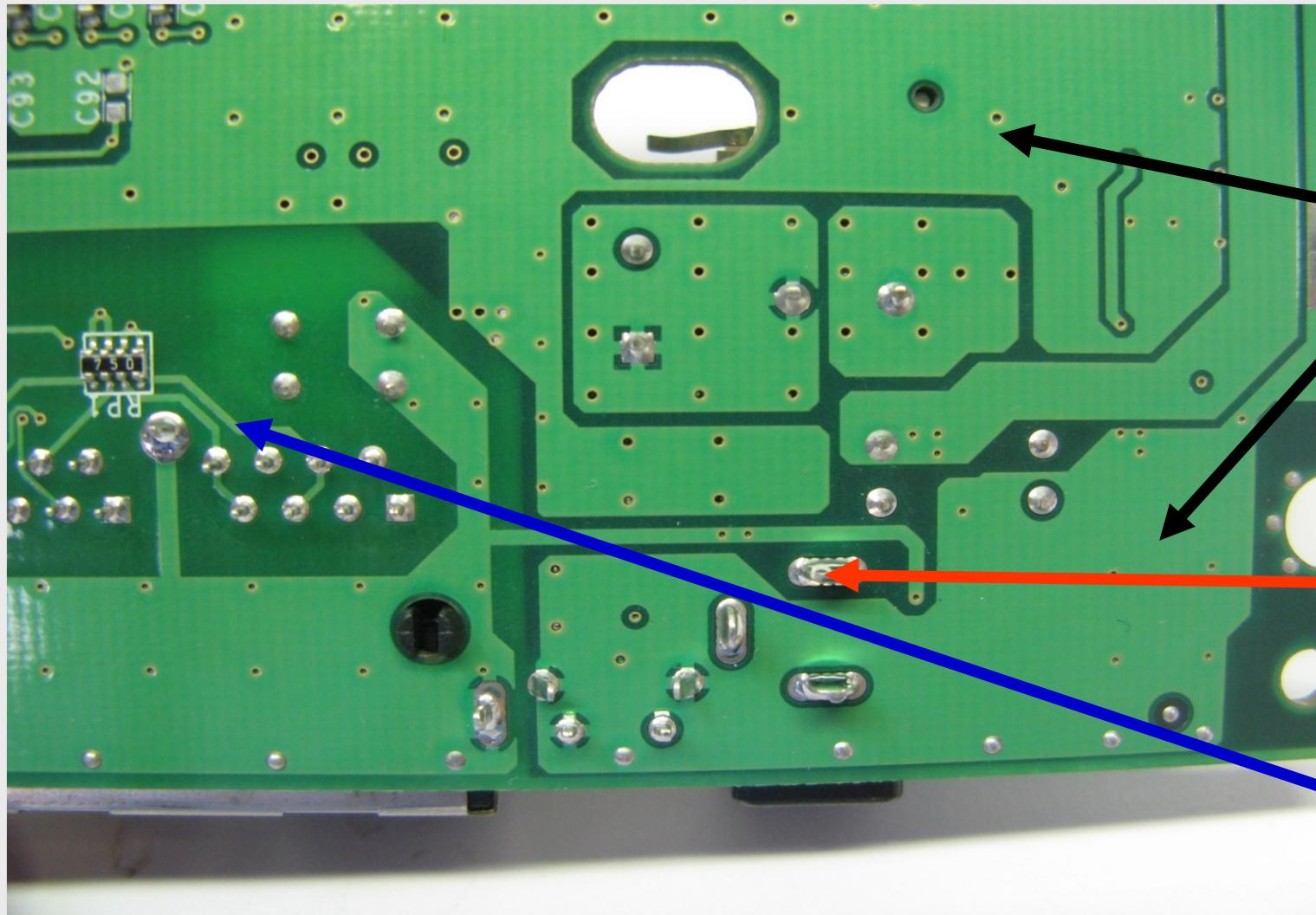
The logarithmic ratio of input power to output power, which, describes the signal attenuation along a defined transmission path.



- System attenuation  $A = 20 \cdot \log \frac{Z_A + Z_F + Z_B}{Z_A + Z_B} \text{ in } (dB)$

- Impedance  $Z_F = \left[ 10^{\frac{A}{20}} \cdot (Z_A + Z_B) \right] - (Z_A + Z_B) \text{ in } (\Omega)$

# Insertion loss calculation



# Insertion loss calculation

$$Z_F = \left[ 10^{\frac{A}{20}} \cdot (Z_A + Z_B) \right] - (Z_A + Z_B)$$

$$Z_F = \left[ 10^{\frac{12}{20}} \cdot (10_A + 10_B) \right] - (10_A + 10_B)$$

$$Z_F = 59.6 \Omega$$

$$A = 20 \log \frac{Z_A + Z_F + Z_B}{Z_A + Z_B}$$

$$A = 20 \log \frac{10_A + 59.6_F + 10_B}{10_A + 10_B}$$

$$A = 20 \log 3.98$$

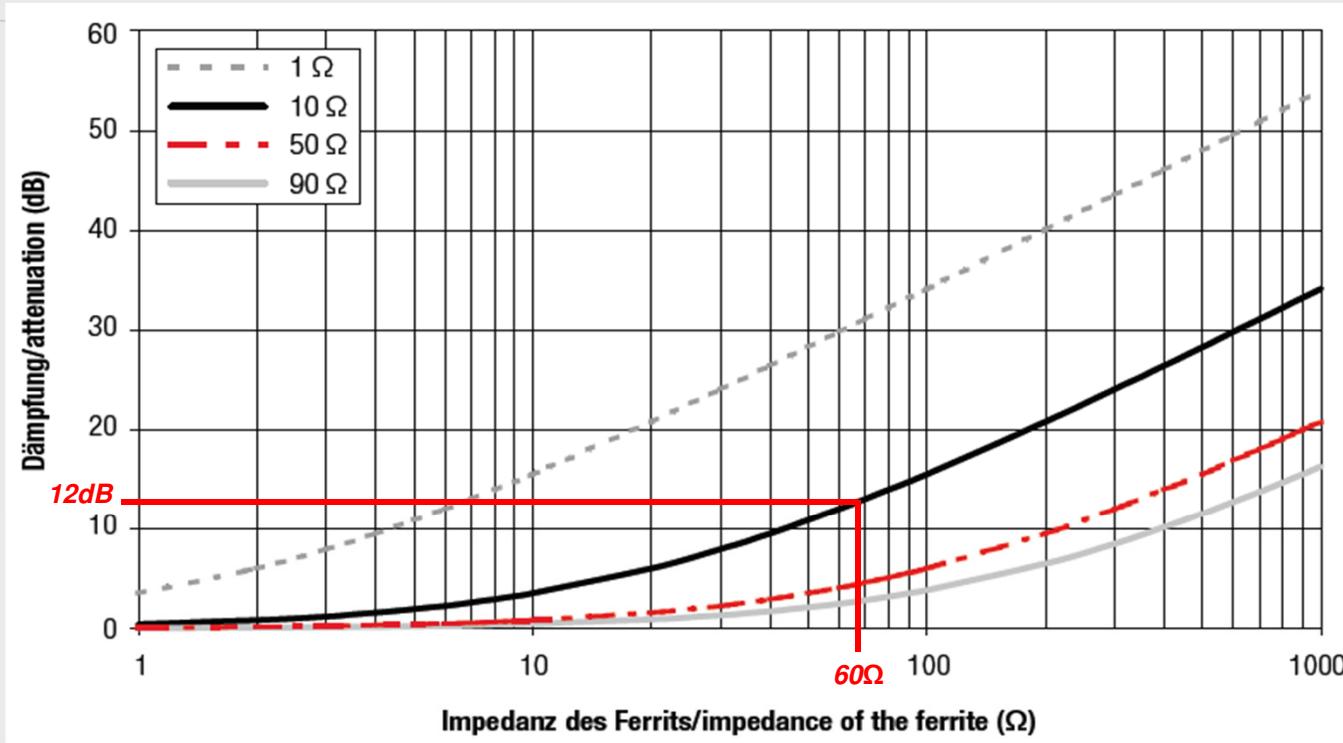
$$A = 11.99 \text{ dB}$$

1. Require 12dB of attenuation at 125 MHz
2. Know that it is a power cable
3. Power port has 10 Ω impedance
4. Result is a impedance of 60 Ω

## Application overview

Assumed practical system impedance	Application
1 Ω	GND (Ground Planes)
10 Ω	V <sub>cc</sub> (Supply Voltage lines)
50 Ω – 90 Ω	Datasignal Lines/Clock/Video Signal/USB
90 Ω – 150 Ω	Long Datasignal Lines

# Insertion loss calculation



## Application overview

Assumed practical system impedance	Application
1 Ω	GND (Ground Planes)
10 Ω	$V_{cc}$ (Supply Voltage lines)
50 Ω – 90 Ω	Datasignal Lines/Clock/ Video Signal/USB
90 Ω – 150 Ω	Long Datasignal Lines

1. Require 12dB of attenuation at 125 MHz
2. Know that it is a power cable
3. Power port has 10 Ω impedance
4. Result is a impedance of 60 Ω

Errors are generally due to wrong system impedance estimation

→ Too much attenuation. Reduce the impedance of ferrite

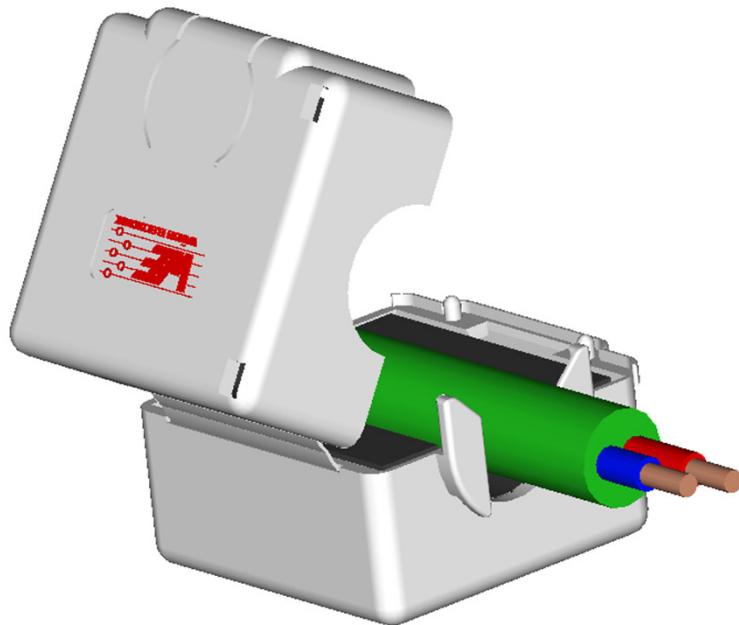
→ Too little attenuation. Increase the impedance of ferrite

## CLAMP-ON FERRITES

# Clamp On Ferrites



Comparable Performance to a bifilar winding Common Mode Choke



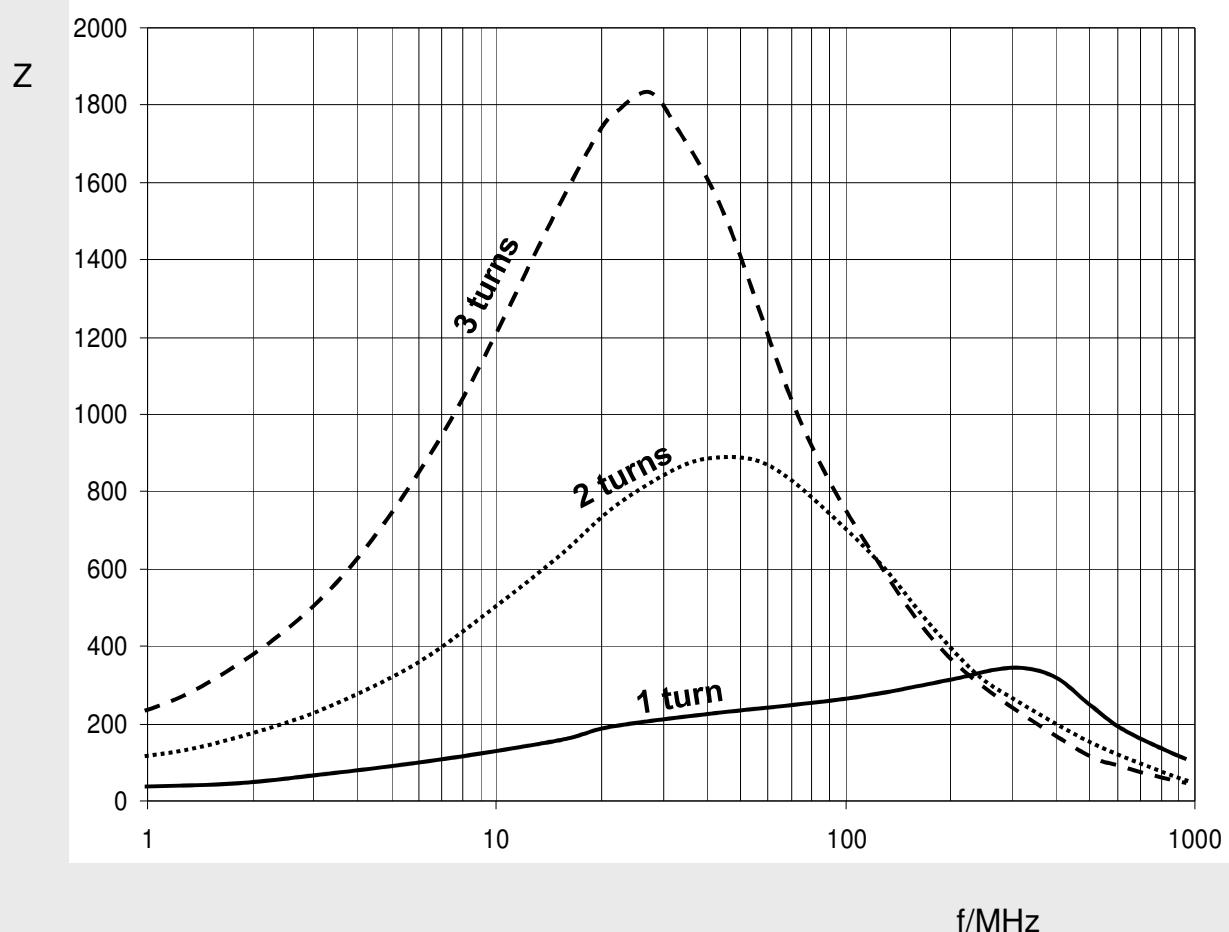
~



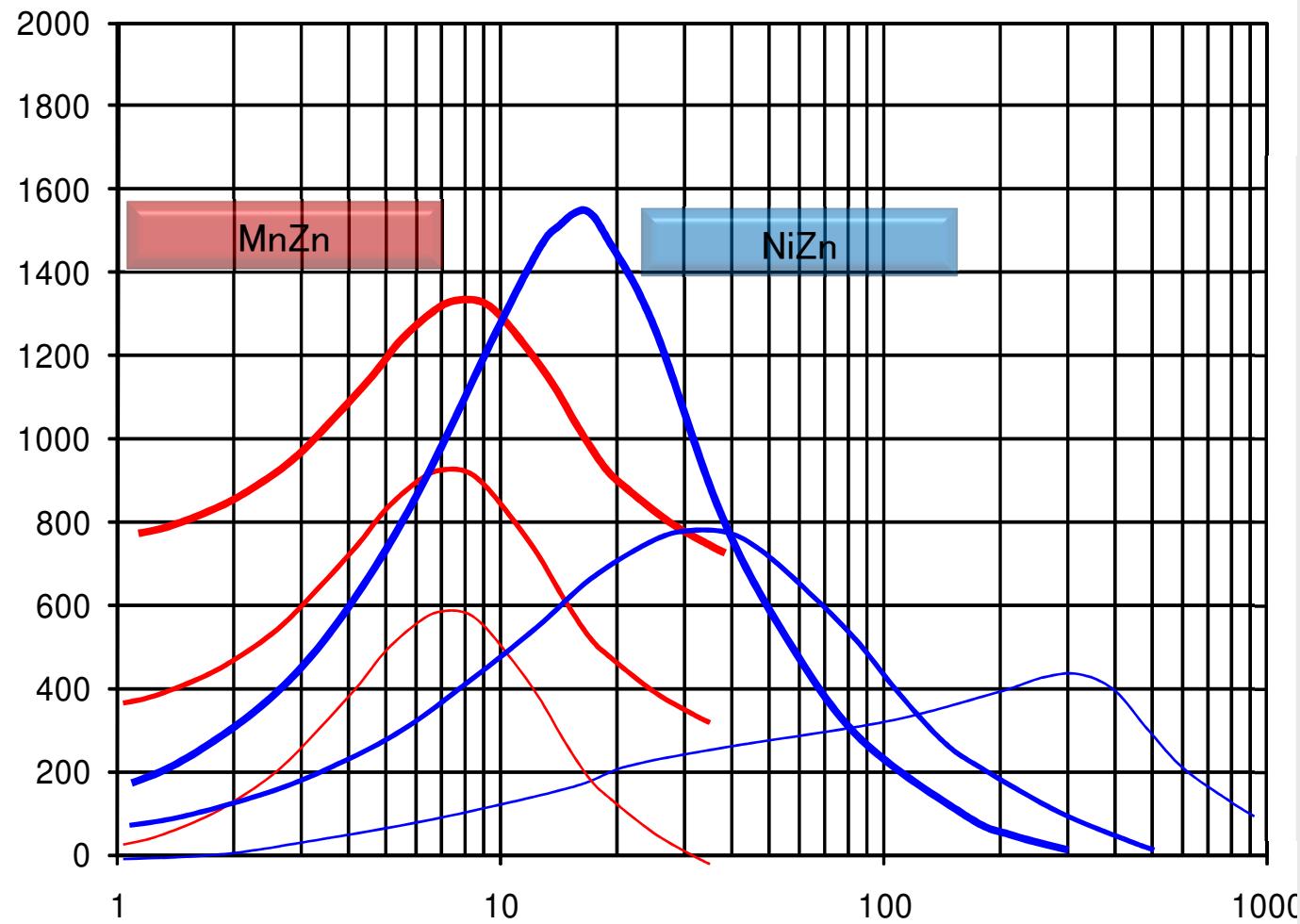
- Both will absorb Common Mode interferences
- Clamp on Ferrite is a Common Mode Choke with ONE winding

# Clamp on Ferrites – Differential mode noise

- Can also be used as a differential mode filter.
  - In this example, it is single wire.

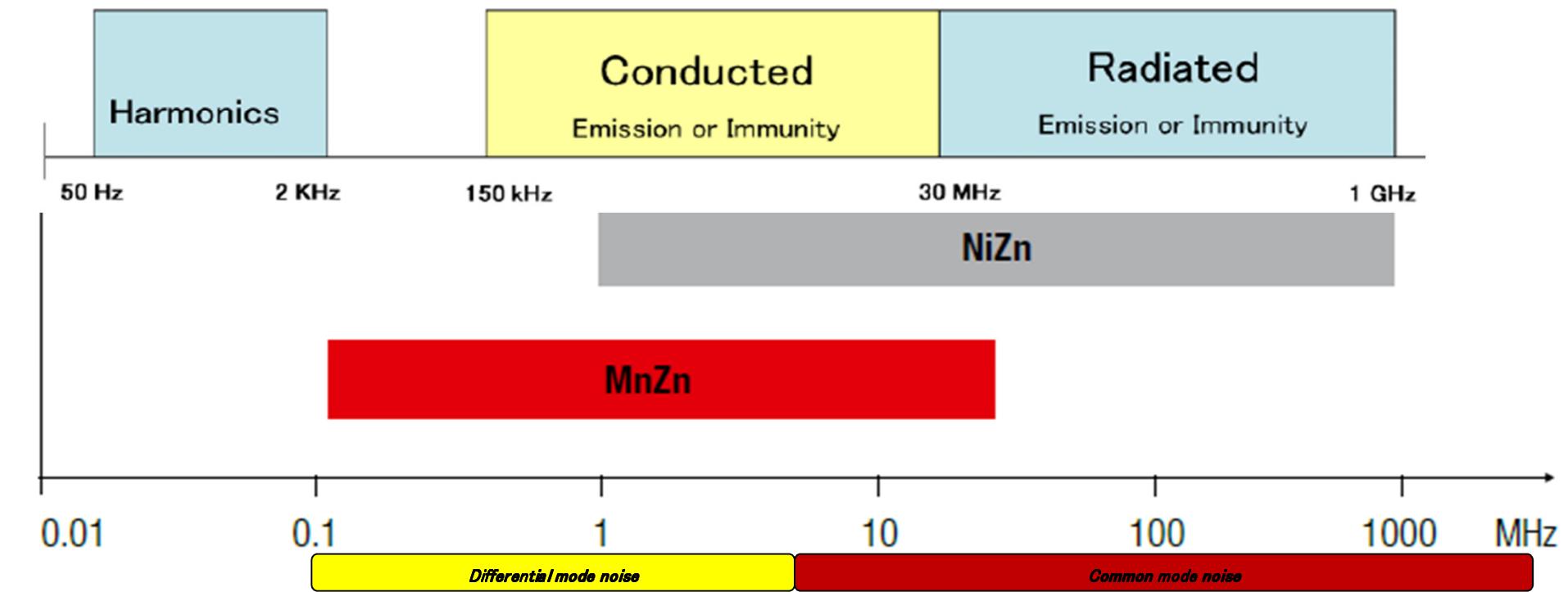


# Clamp on Ferrites. Material Comparison



# Clamp on Ferrites

## Material characteristics



# How can we find out what interference we have?

## Common mode or differential mode?

Take a Snap Ferrite and fix it on the cable  
(both lines e.g. VCC and GND)

if noise is reduced or  
noise immunity  
is increased

Common Mode interferences

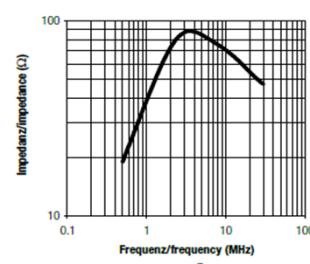
if not

Differential Mode interferences

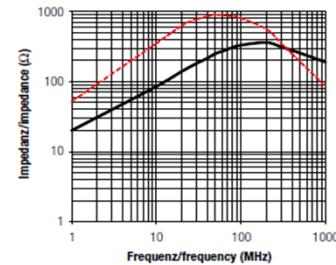
# Clamp on Ferrites



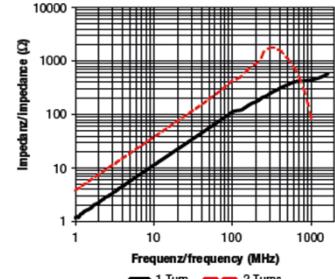
Impedance vs. Frequency



742 711 12 / 742 711 12S



742 716 33 / 742 716 33S



FIX – LFS

TEC / RING / FIX

GAP

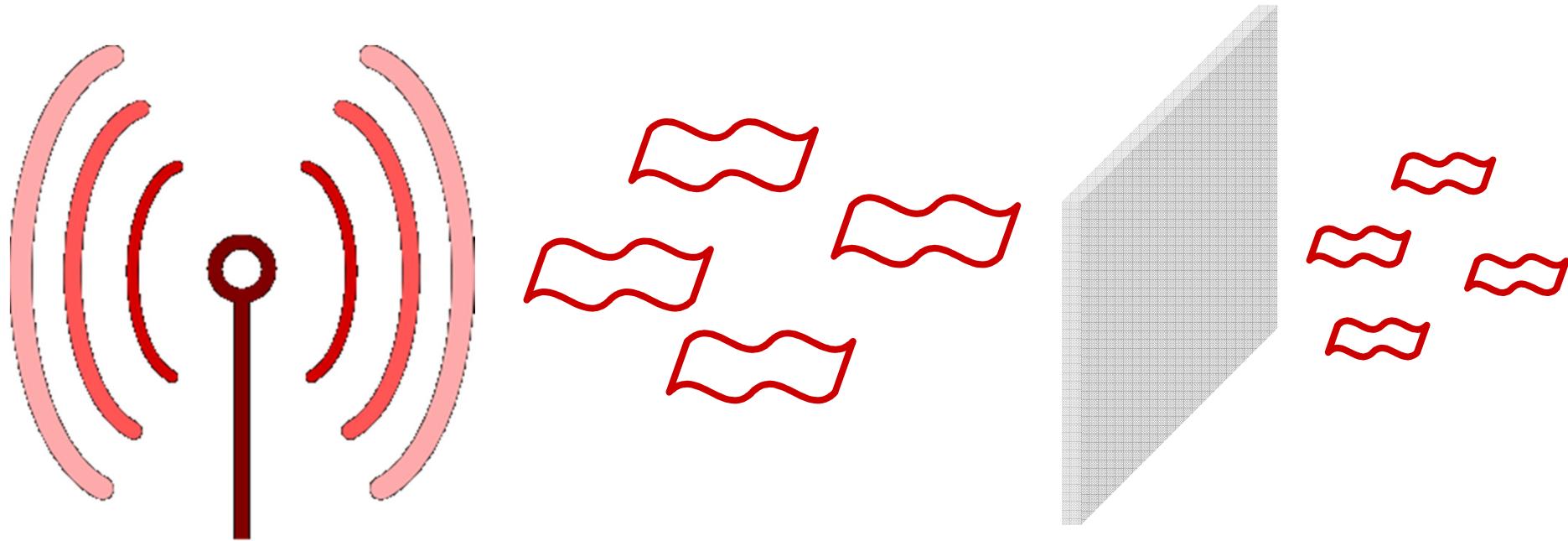
## • STAR-GAP

- Developed to increase impedance and to lower the suppression effect in the high frequency range with DC bias.
  - Best performance with two turns.
  - Very helpful in lowering EMI problems in frequency range of 100MHz up to 2.5GHz.

# SHIELDS

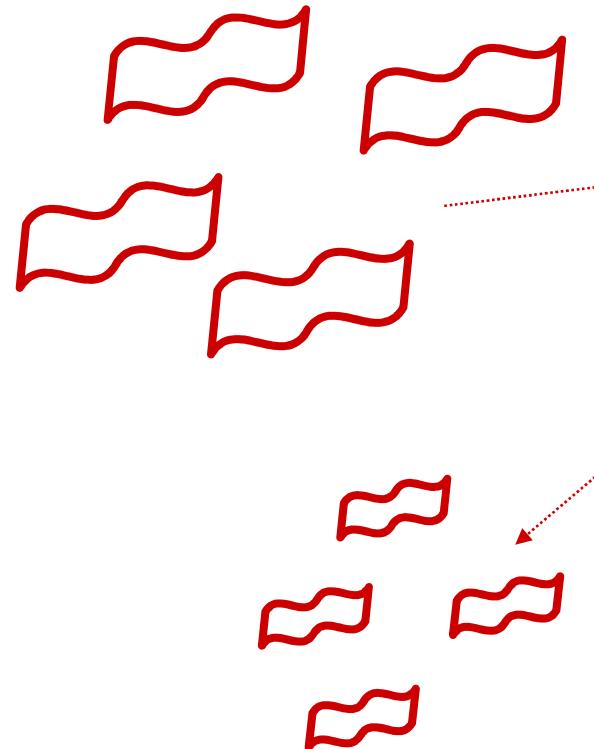
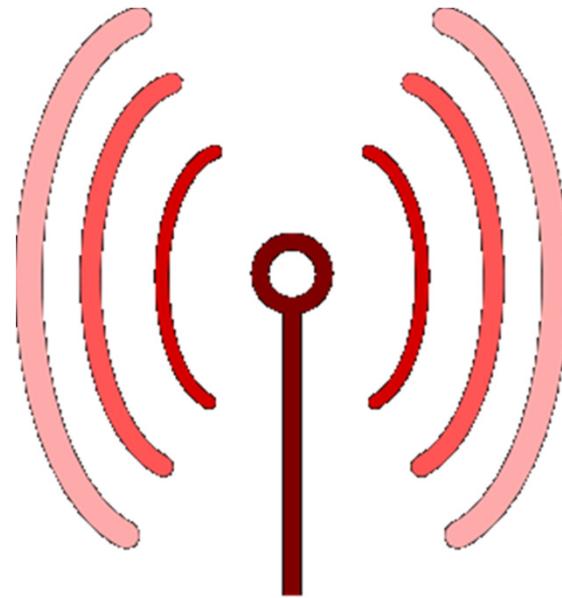
# Radiated noise modes

Transmission of radiated noise



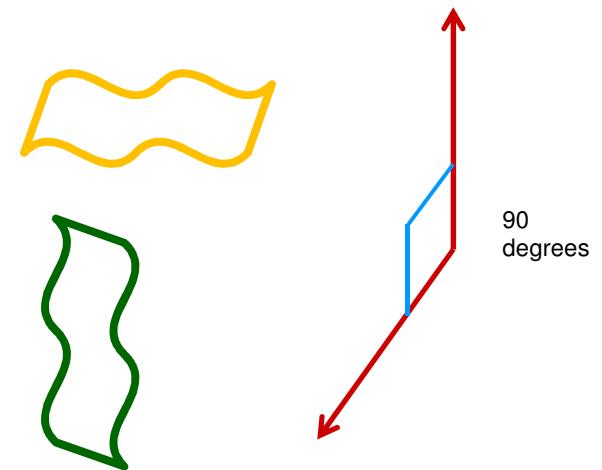
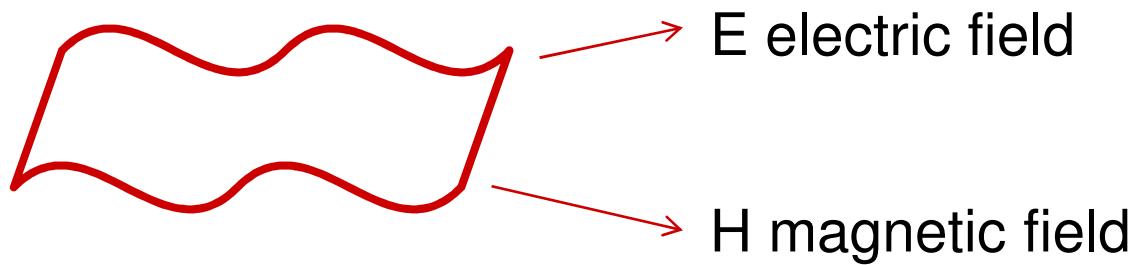
# Radiated noise modes

## Reflection of radiated noise

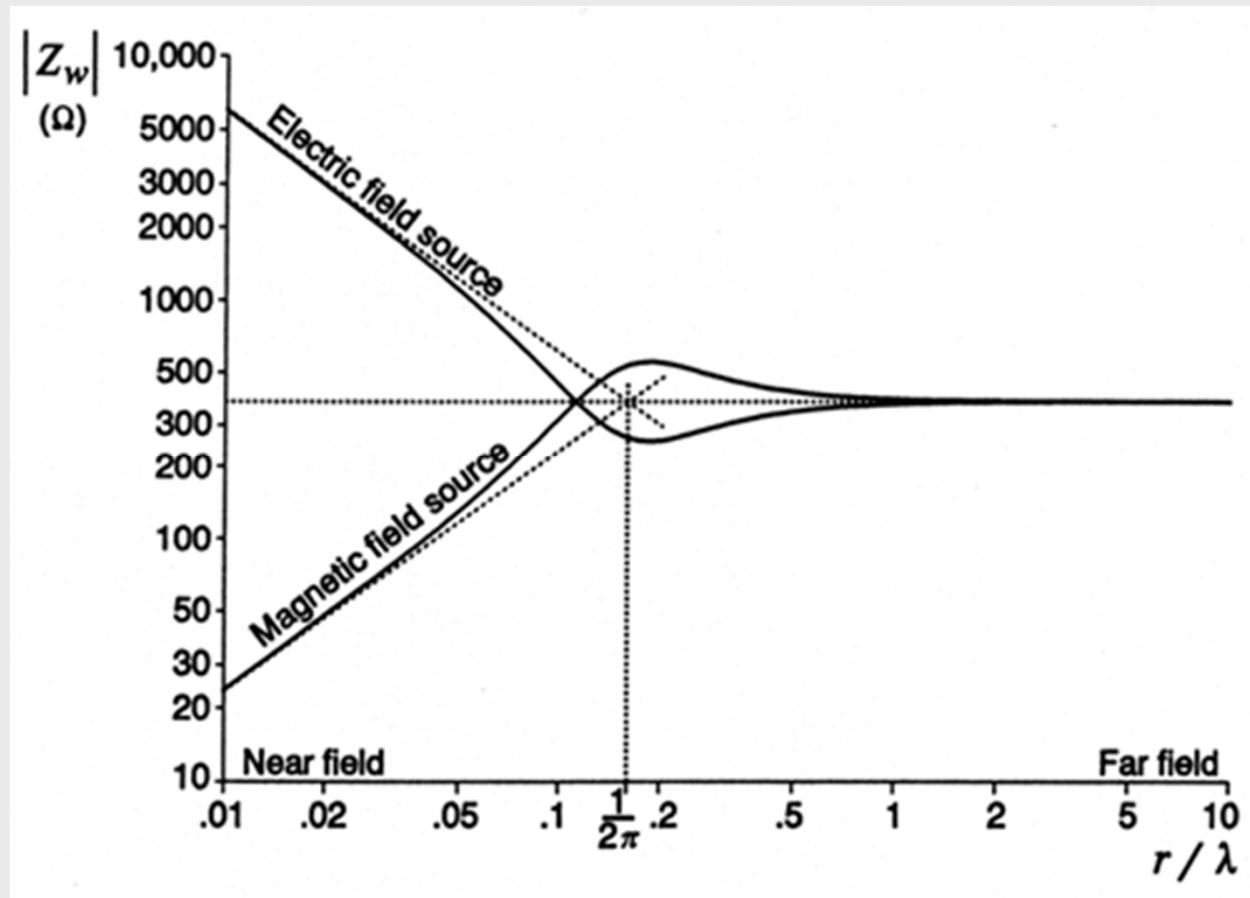


# Electromagnetic wave

Components from a magnetic wave



# Near Field and Far Field



# Enclosures

- Products can be built with two materials regarding RF:

**1. Metallic – Aluminium, steel, brass, copper.**

**Natural shielding materials**

**2. Non metallic – Plastic, Nylon, Polystyrene, PVC.**

**Materiais transparent to RF**



# RF shielding types



Principal two type: **Metallic** (Cu, Al, Steel ) and **Ferrite** (NiZn, MnZn or similar compounds)



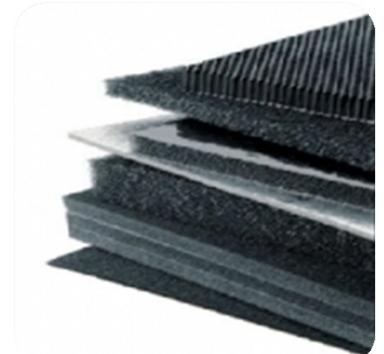
**metallic shielding**  
**WE-SHC**



**Absorber sheet**  
**WE-FAS**



**Conductive gasket**  
**WE-LTS**



**Conductive foam**  
**WE-LS**



**Conductive gasket**  
**WE-LT**

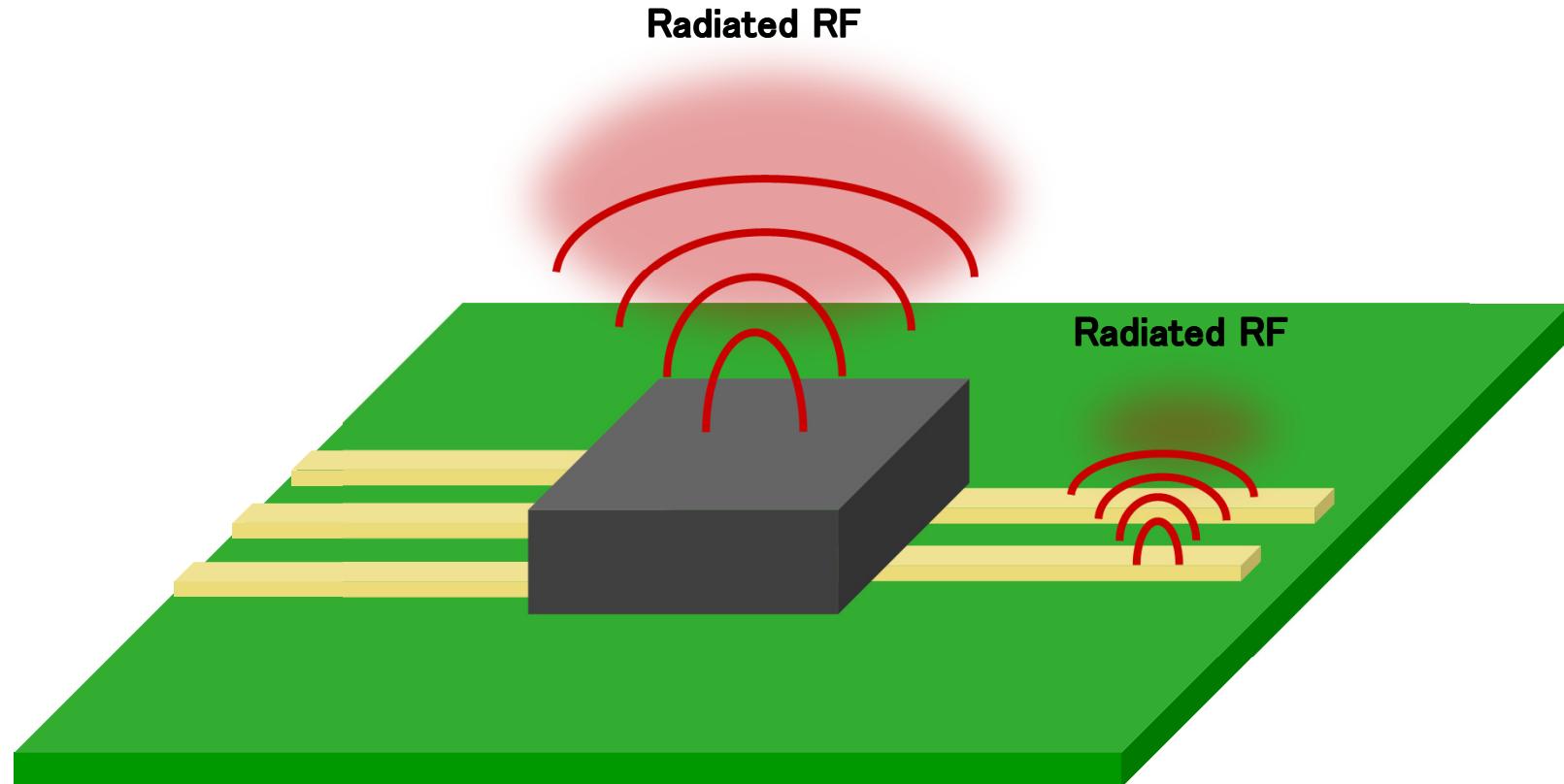


**Conductive tapes**  
**WE-CF/TS**



**Brooke Shields**

# Basic problem to be solved



# Applications (Shielding)



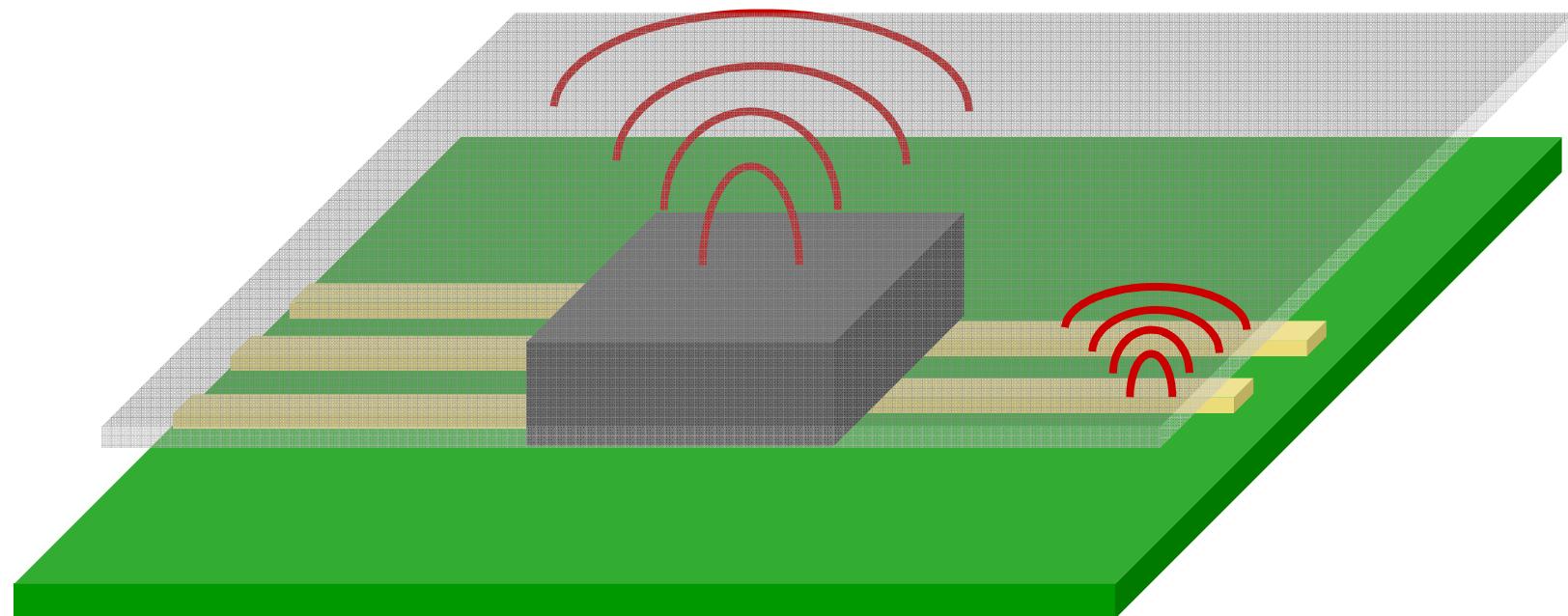
metallic



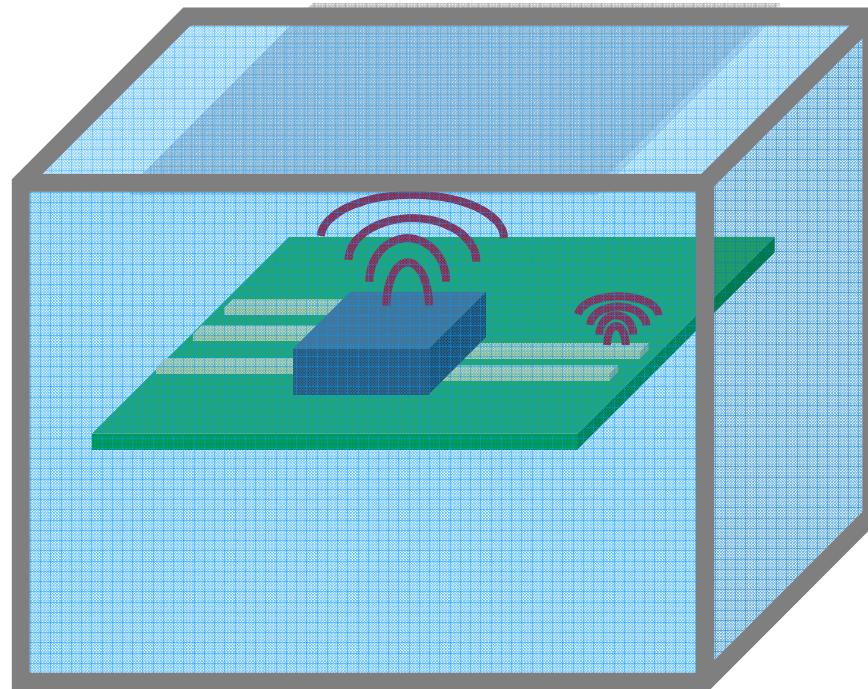
Ferrite



# Application ( Shielding components )

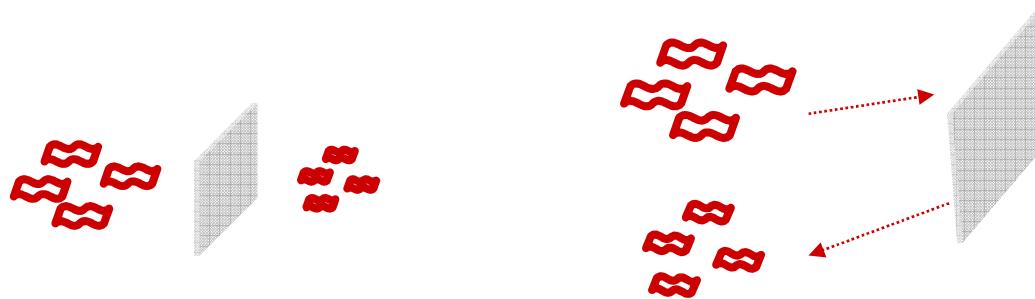


# Application ( shielding enclosures )

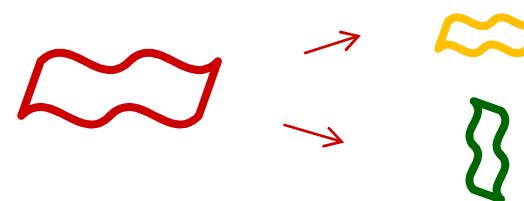


# Revision

Two paths: transmission and reflection



Two components E field and H Field



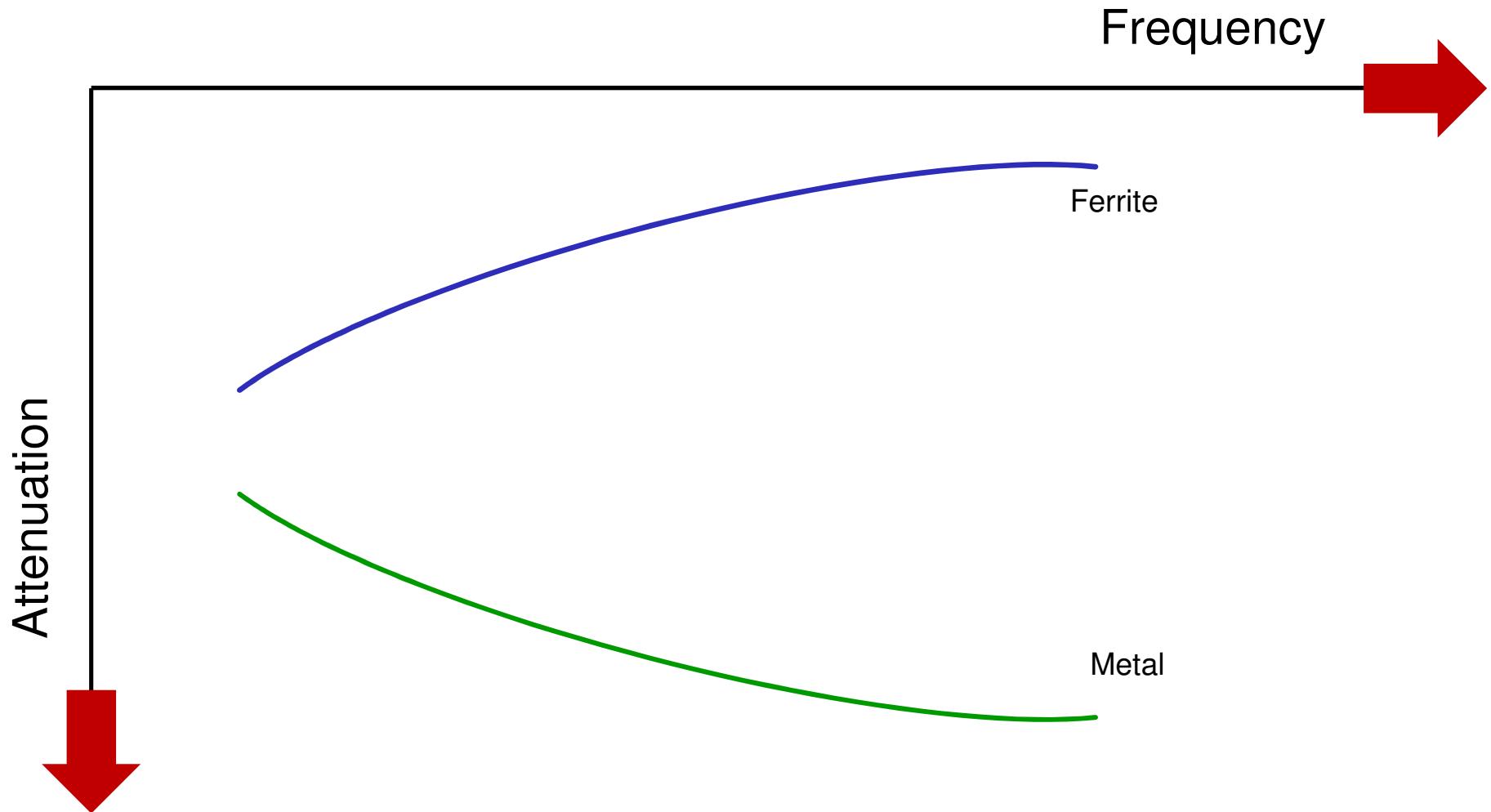
Two types of material



# Material Behavior with Radiated Signals



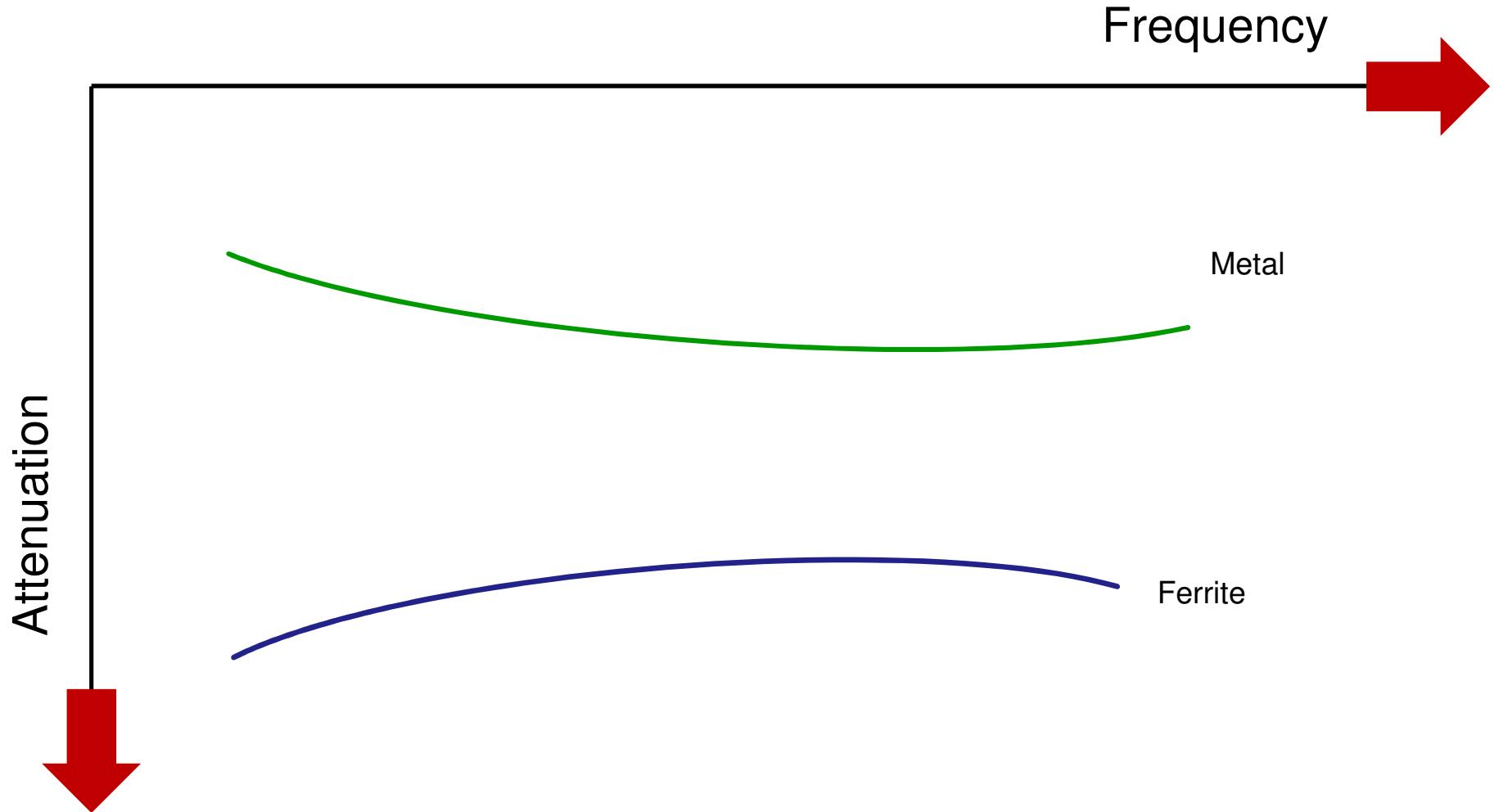
Metal vs Ferrite when signals are transmitted



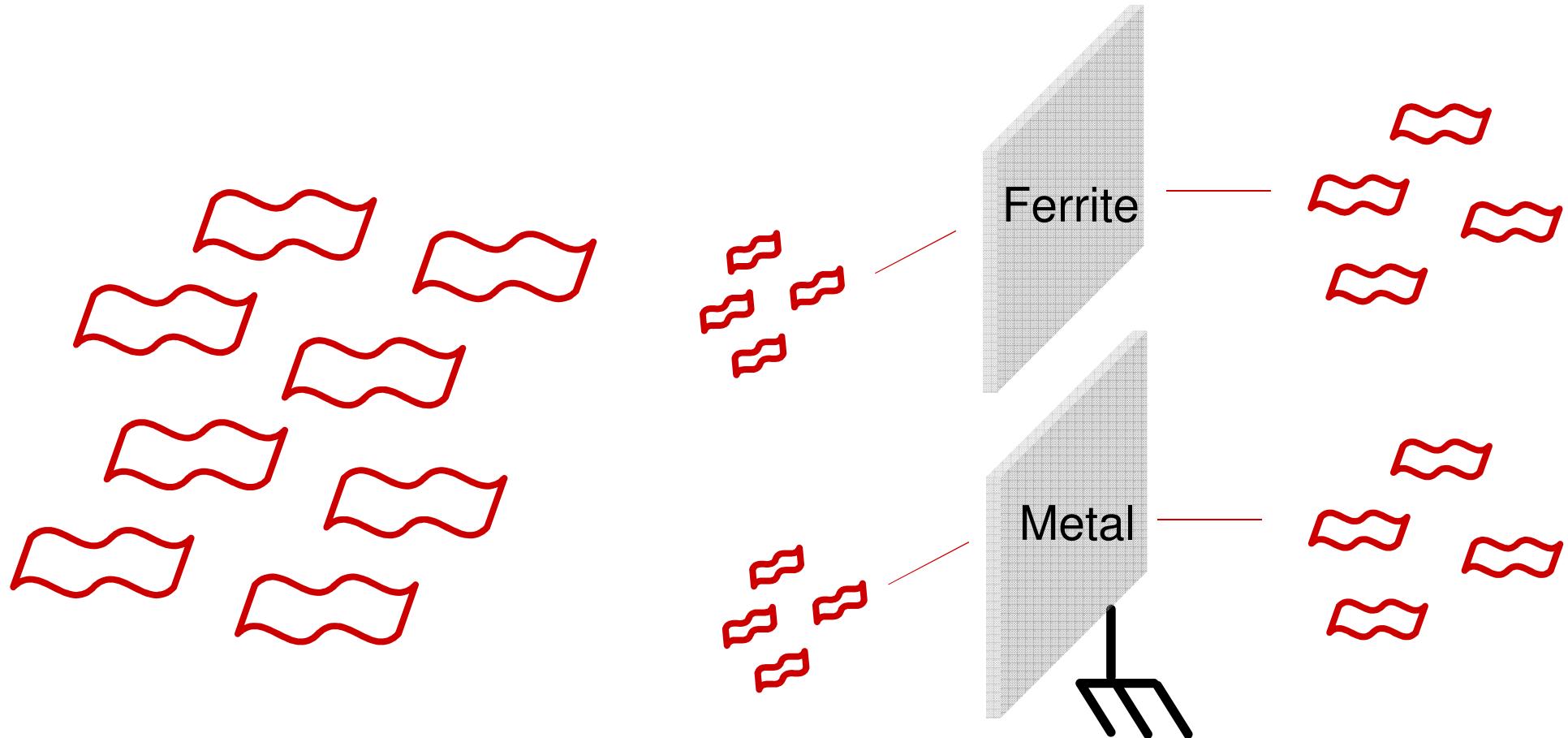
# Material Behavior with Radiated Signals



Metal vs Ferrite when signals are reflected



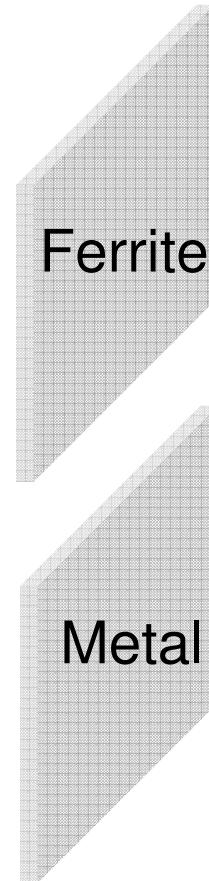
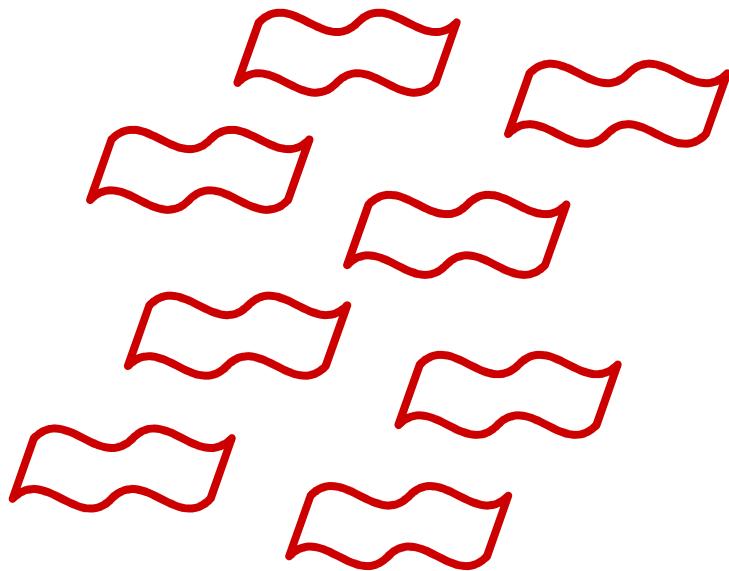
# Advantages Metal vs Ferrite



Metal has to be grounded! But a ferrite does not.

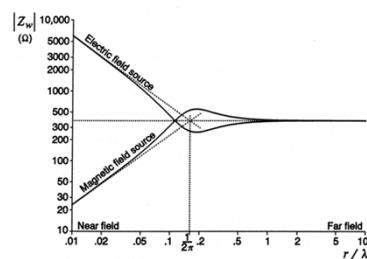
Metal is cheaper than ferrite and easy to find.

# Shield Effect on Components of the EM Wave



Ferrite works better at short distances and it affects more significantly the magnetic field

Metal works at any distance and it affects more significantly the electric field



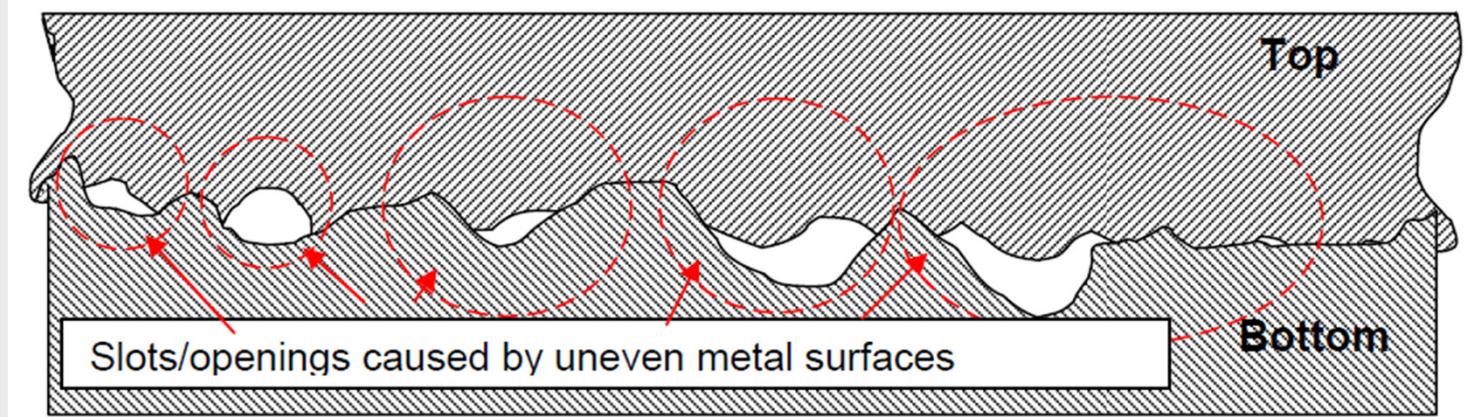
# Applications ( scenario )



## Metallic enclosure



# Problems with metallic enclosures



- **Gaps allow RF noises pass through.**
- **Slots can work as antennas**

Antena WLAN 2,4GHz

$$\lambda = \frac{3 \times 10^8 \cdot m \cdot s}{2,4 \times 10^9 \cdot s}$$

$$\lambda = 0,125m$$



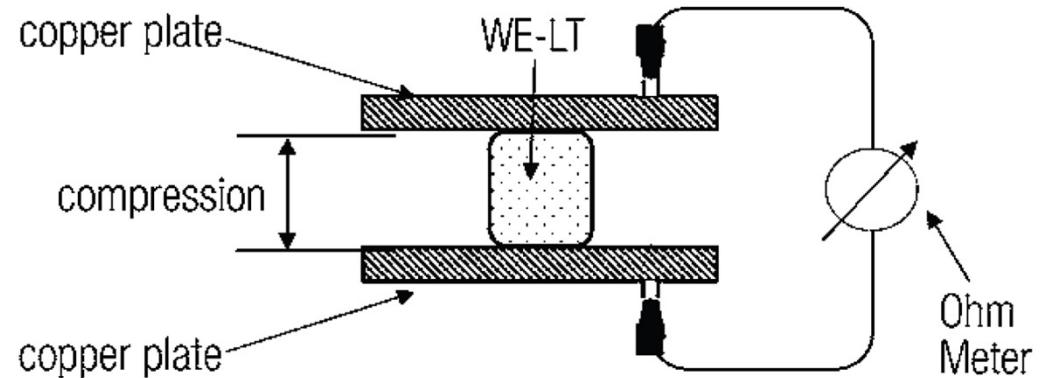
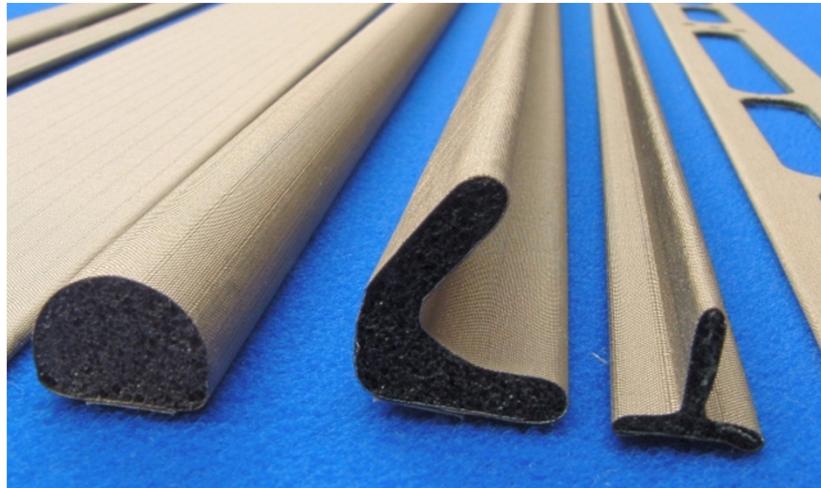
$$\lambda / 4 = 3,125cm$$



# Metallic enclosures correctly grounded



## Gaskets and conductive tapes

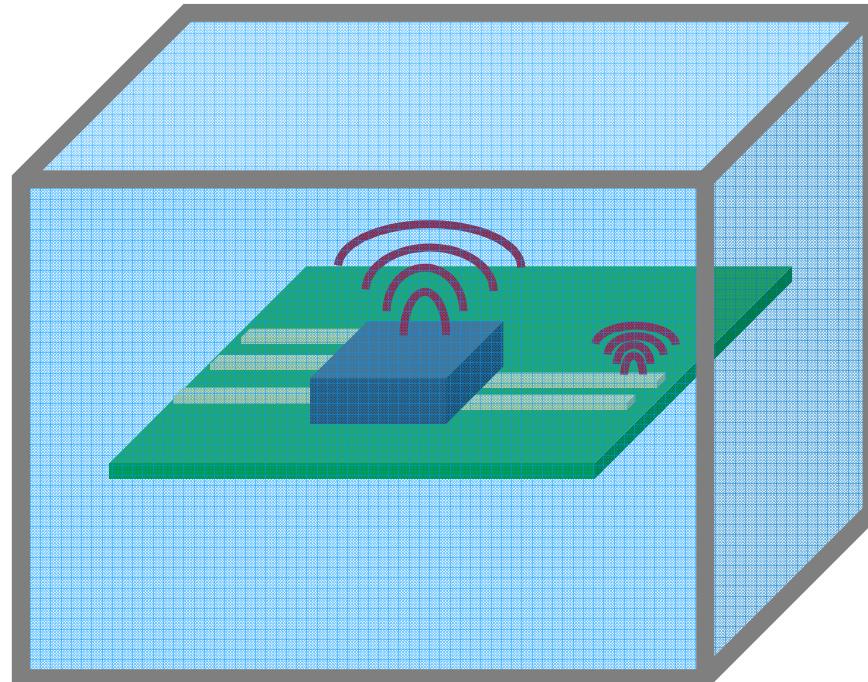


Best gasket performance is when compressed by 50% to lowest impedance.

# Applications (Gaskets)

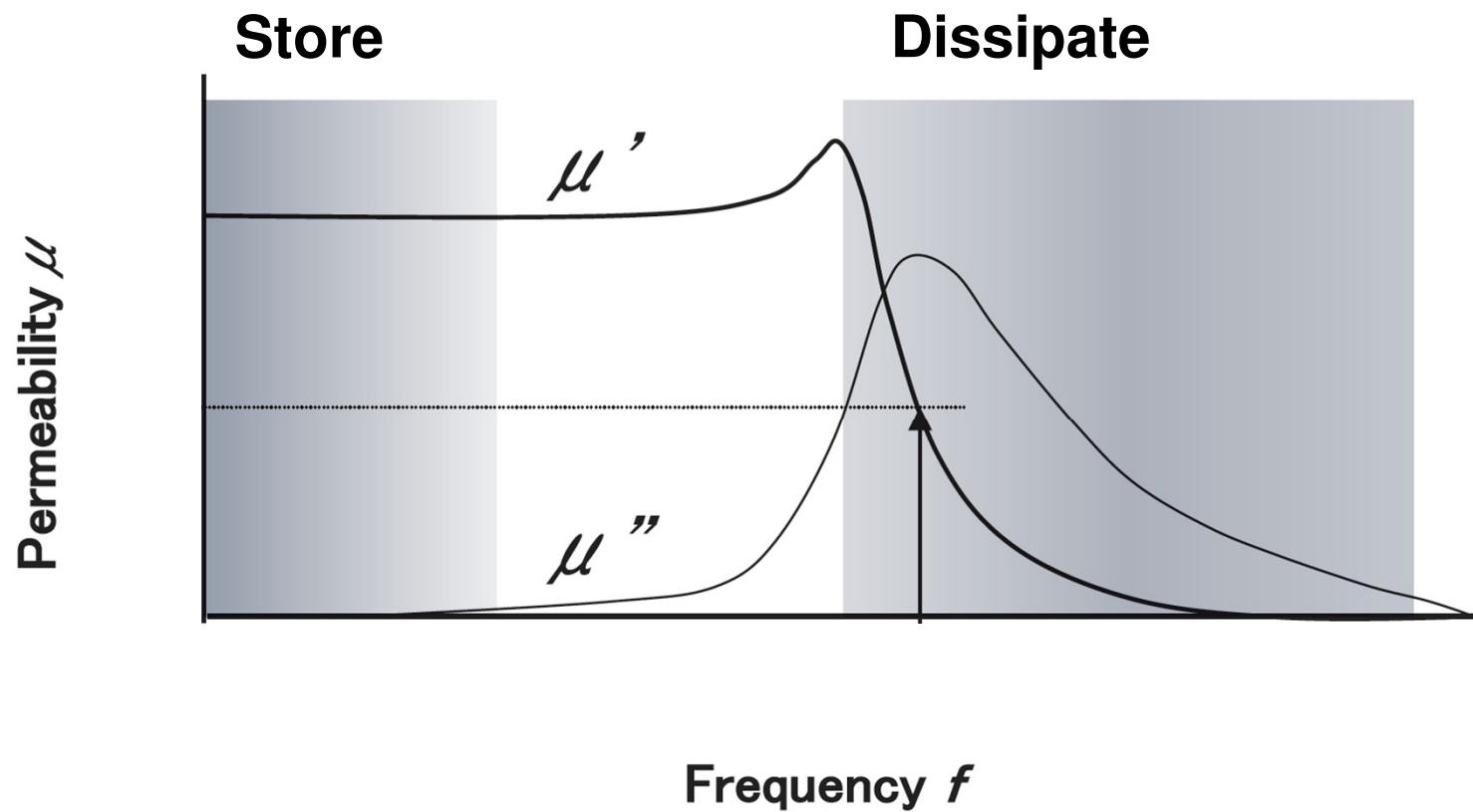


Gaskets fill the gap between metallic walls from enclosure.



## Ferrites characteristics:

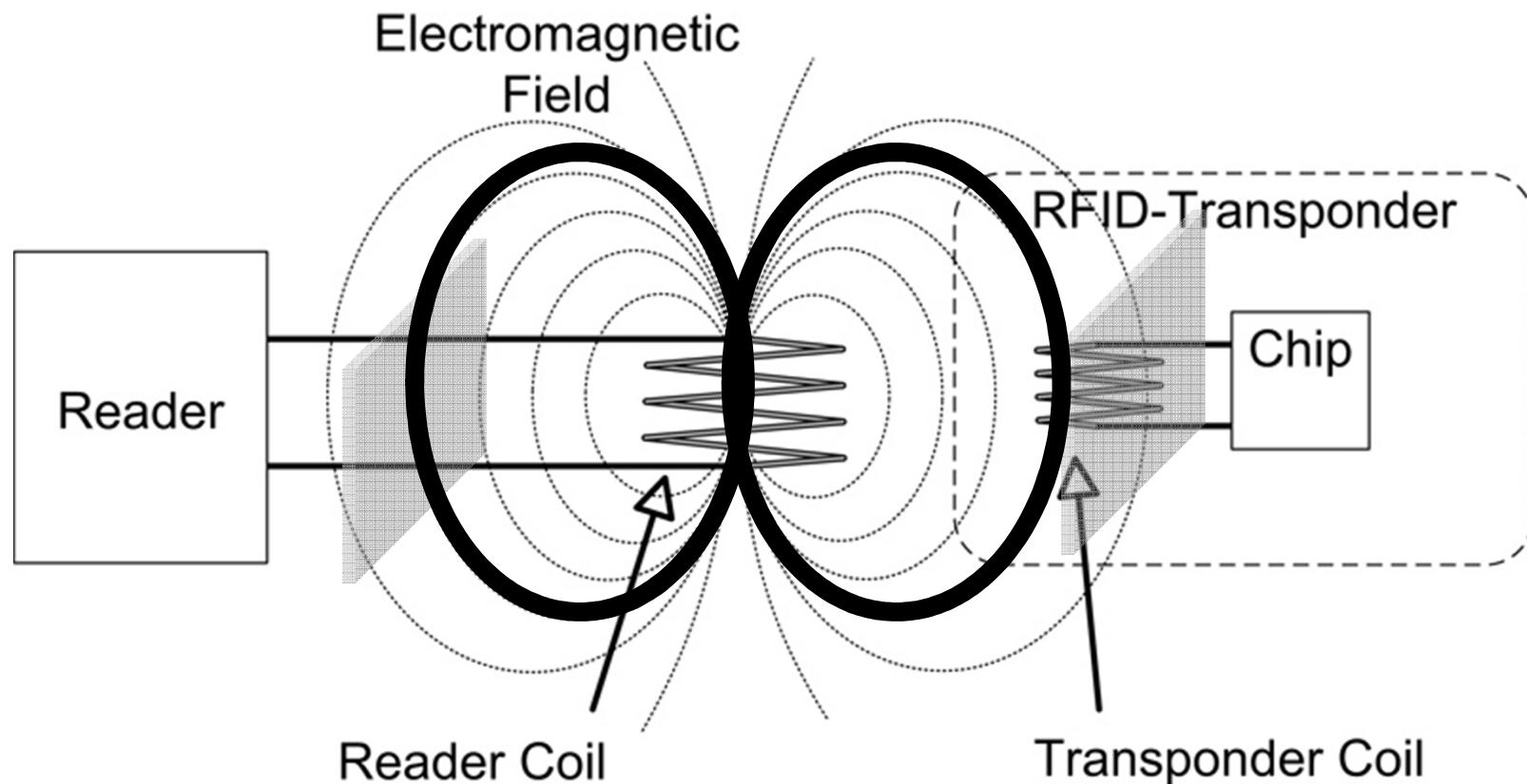
Ferrites can store and dissipate



# Application ( Enhance RFID 13.56MHz)



# Applications ( Enhance RFID 13.56MHz)

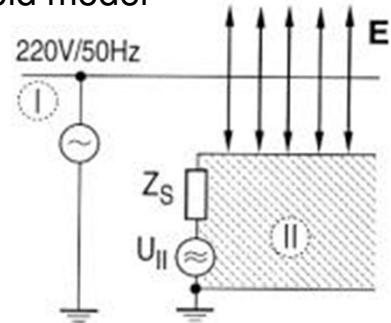


# What is best? Metal or Ferrite?



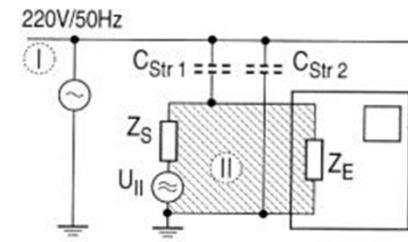
## Capacitive coupling

Field model



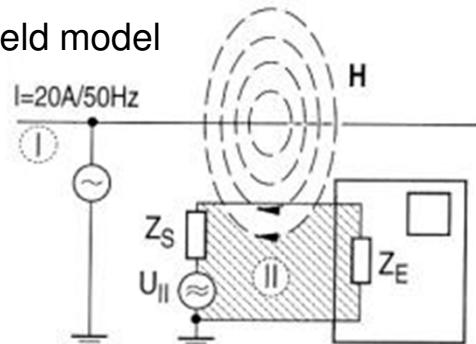
None

Network model



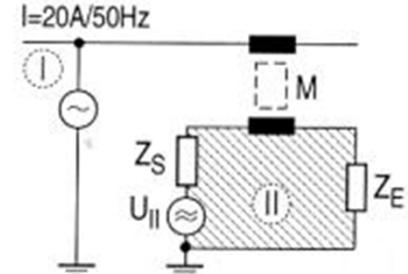
## Inductive coupling

Field model



Ferrite

Network model





# Q&A

<http://www.we-online.com/>

[http://www.we-online.com/web/en/passive\\_bauelemente\\_standard/toolbox\\_pbs/Toolbox.php](http://www.we-online.com/web/en/passive_bauelemente_standard/toolbox_pbs/Toolbox.php)

## **Wurth Electronics Midcom Inc. Headquarters**

Phone: (605) 886-4385

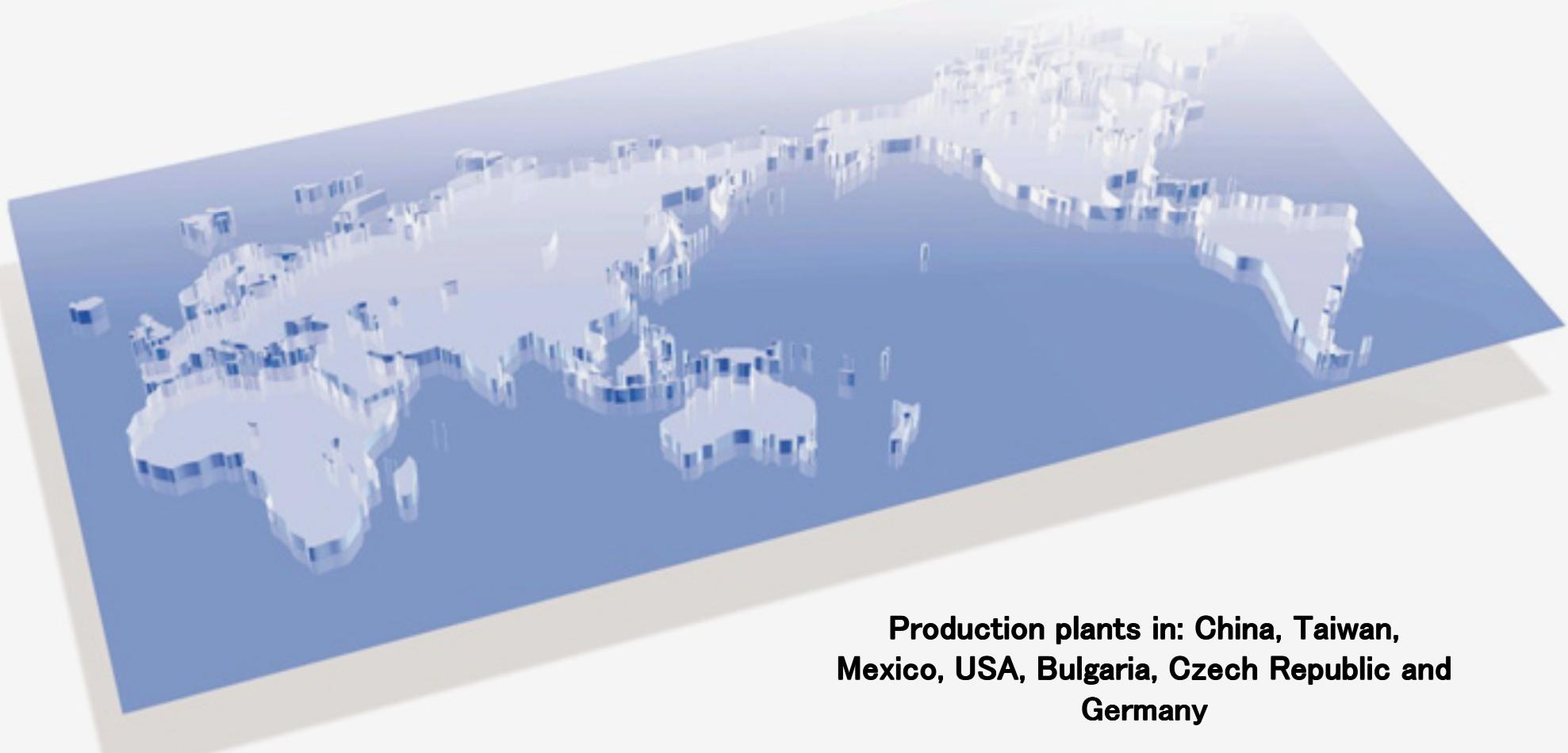
Fax: (605) 886-4486

E-Mail: [midcom@we-online.com](mailto:midcom@we-online.com)

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## APPENDIX

# EMC Overview

## Conductive coupling

- Coupling path between source and victim is formed by direct contact.  
Example a transmission line, wire, cable, PCB trace or metal enclosure.

## Capacitive coupling

- Electric field coupling.

Occurs when a varying electrical field exists between two adjacent conductors.

## Inductive coupling

- Magnetic field coupling

Occurs when a varying magnetic field exists between two parallel conductors.

Coupling between conductors causes parasitic induced voltages.

## Radiative

- Source is the “transmitter” and victim is the “receiver”.

# The Magnetic Field



The H field corresponds to what is called the magnetic field strength. It is measured in amps / meter (A/m).

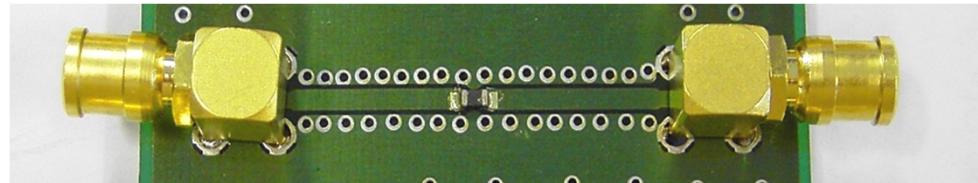
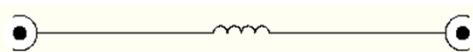
In free space or in air the B field represents magnetic flux density which is given in units of Tesla by  $B = \mu_0 \cdot H$  where  $\mu_0$  is the absolute magnetic permeability of free space  $\mu_0 = 4\pi \times 10^{-7} \frac{N}{A^2}$

More magnetic flux can be produced by the same H value in certain (magnetic) materials, notably iron, and this is accounted by introducing another factor, the relative permeability  $\mu_r$ , giving  $B = \mu_0 \mu_r H$  for magnetic materials.

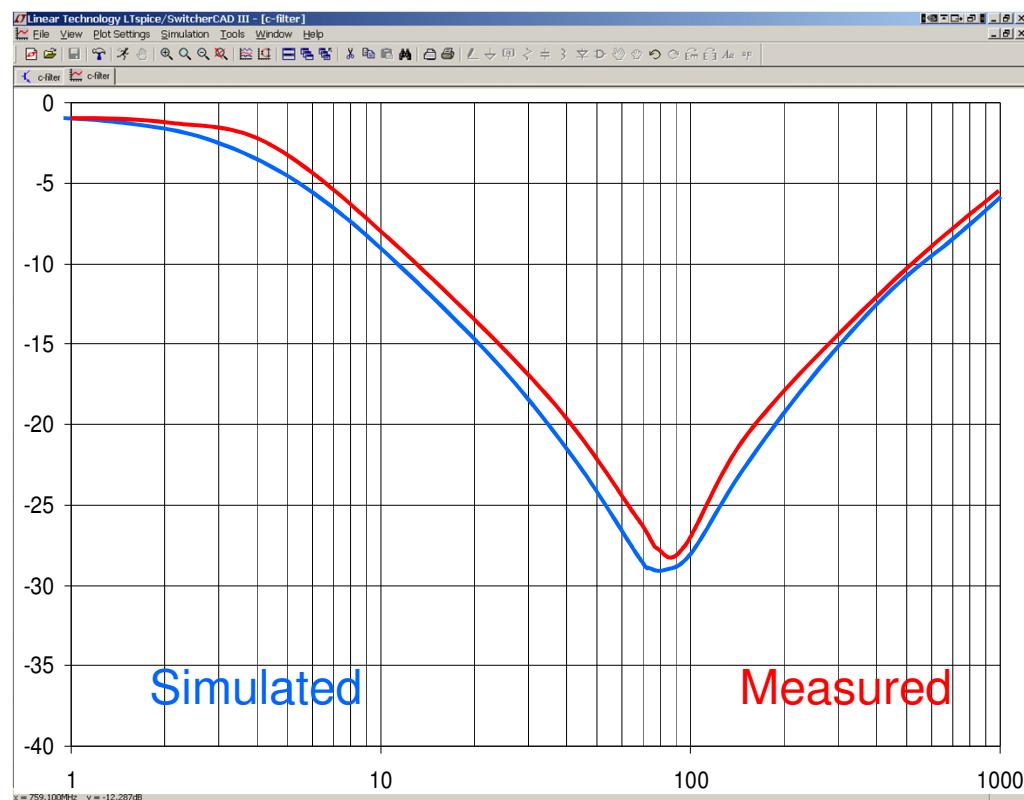
# Filter topologies – L-Filter



- L-Filter



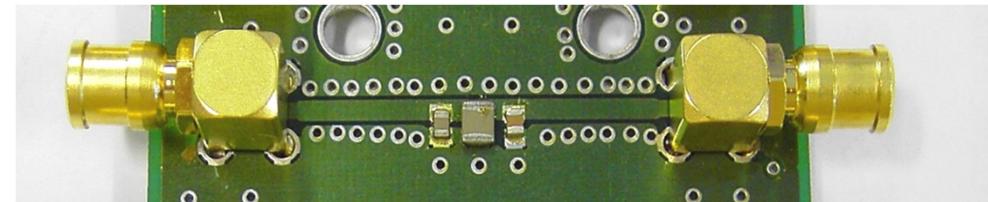
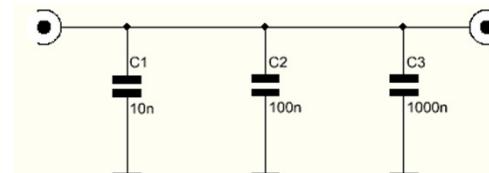
- instead of inductor use a SMD-Ferrite
- WE-CBF 742 792 093
- $Z_{\max} = 3000 \Omega$  @ 80 MHz
- $A_F = -29 \text{ dB}$  @ 80 MHz



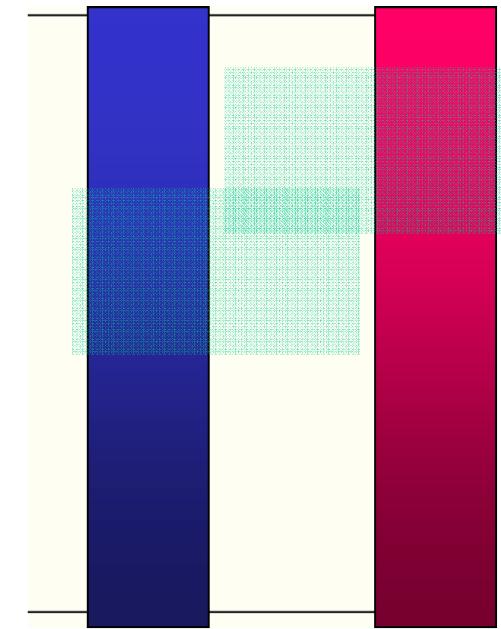
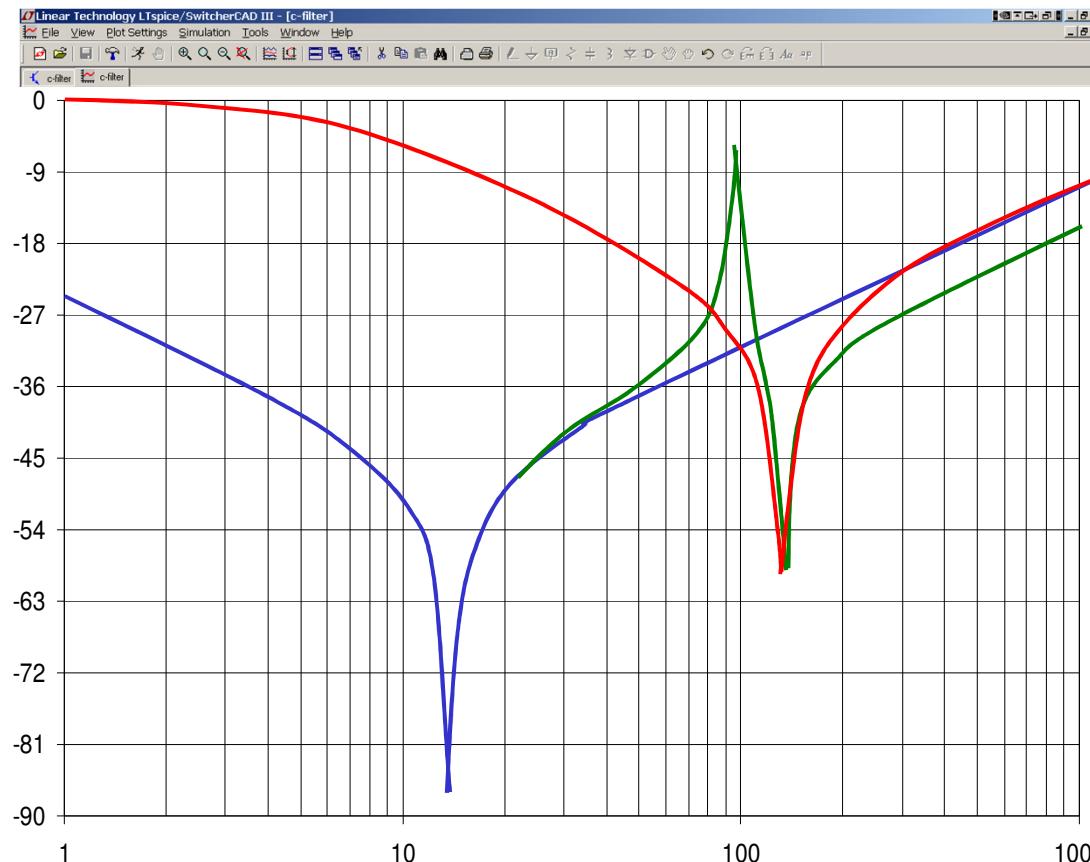
# Filter topologies – Parallel-C-Filter



- Parallel-C-Filter



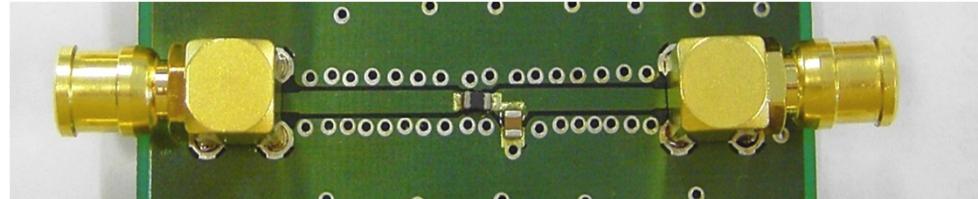
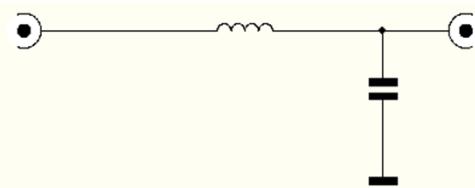
- resonance points



# Filter topologies – LC-Filter



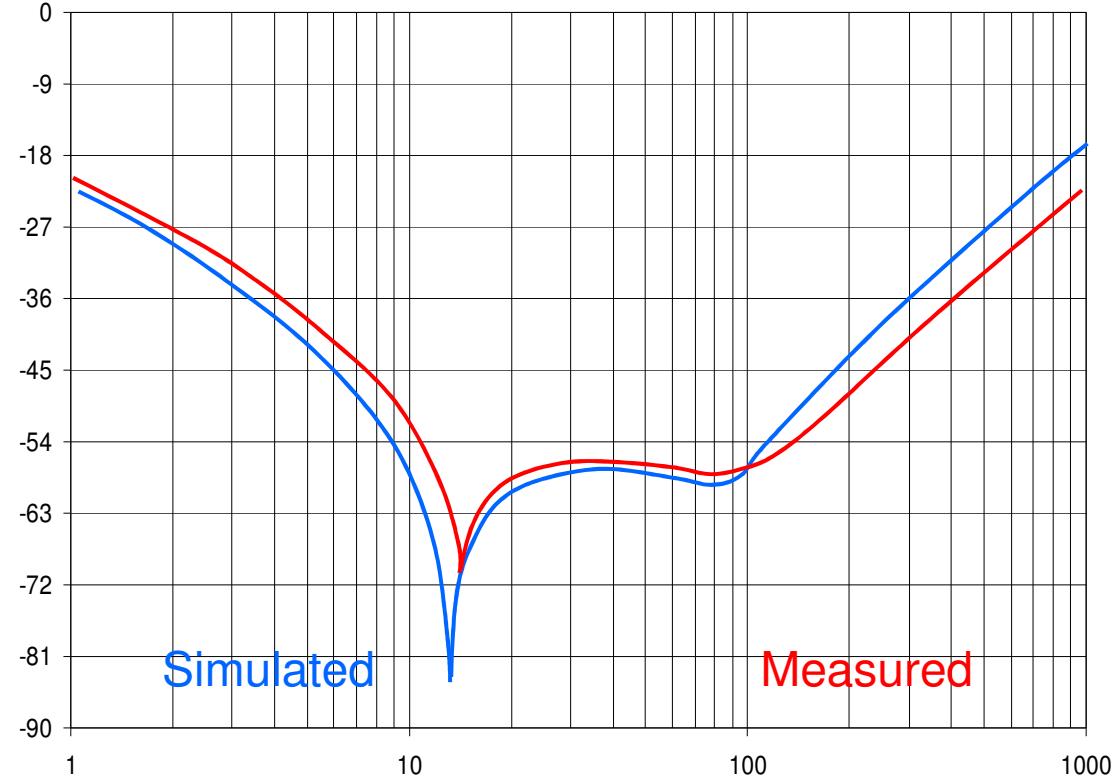
- LC-Filter



- Comparison simulated vs. measured

WE-CBF 742 792 093

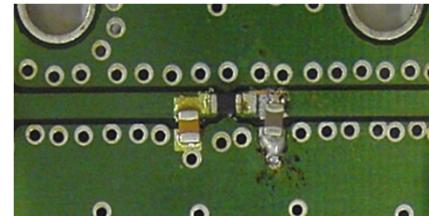
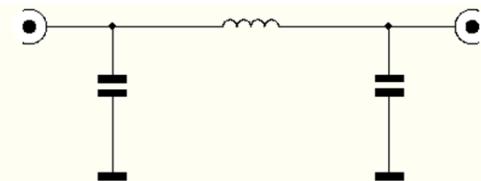
C=100nF



# Filter topologies – PI-Filter



- $\pi$ -Filter

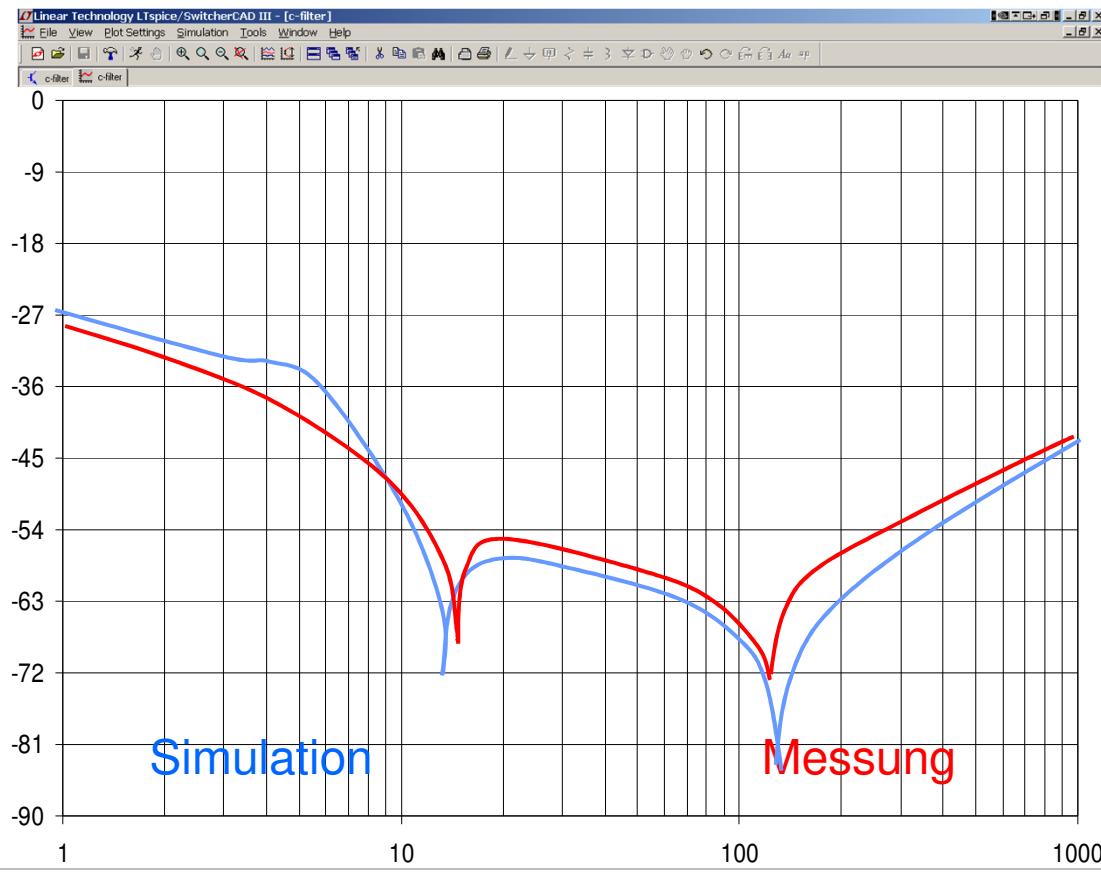


- Comparison Measurement - Simulation

WE-CBF 742 792 093

$C_1 = 1\text{nF}$

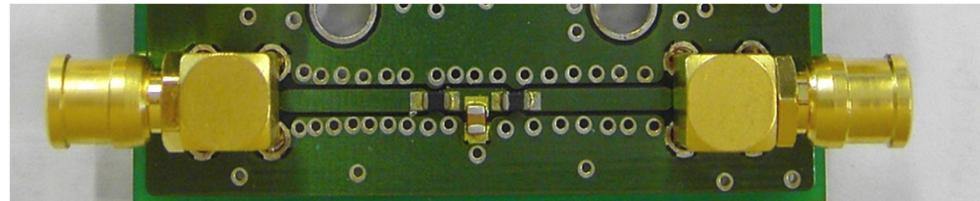
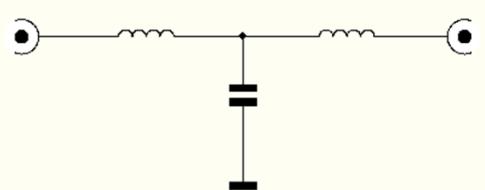
$C_2 = 100\text{nF}$



# Filter topologies – T-Filter



- T-Filter

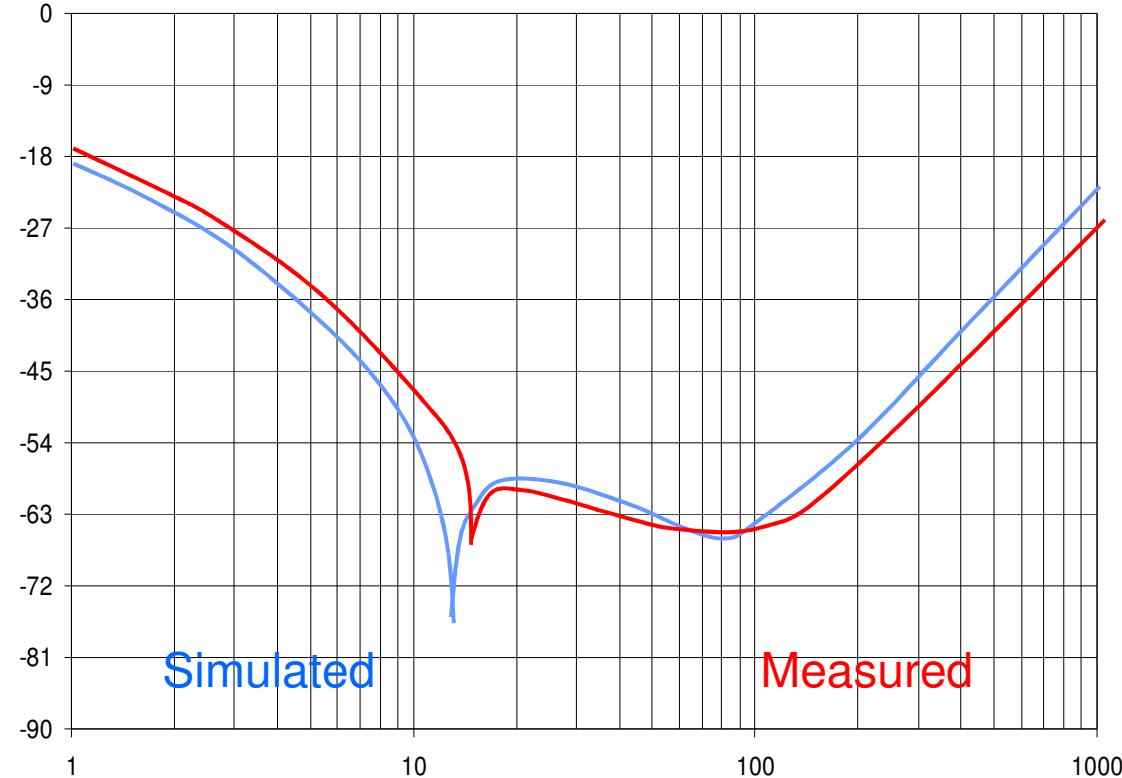


- Comparison simulated vs. measured

$C=100\text{nF}$

$L_1=742\ 792\ 040$

$L_2=742\ 792\ 093$



# Soft ferrites (Typical permeability values)



- Iron power / Superflux : 50 ~ 150
- Nickel Zinc : 40 ~ 1500
- Manganese Zinc : 300 ~ 20000

# The Magnetic Field



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In free space or in air the B field represents magnetic flux density which is given in units of Tesla by  $B = \mu_0 \cdot H$  where  $\mu_0$  is the absolute magnetic permeability of free space  $\mu_0 = 4\pi \times 10^{-7} \frac{N}{A^2}$

More magnetic flux can be produced by the same H value in certain (magnetic) materials, notably iron, and this is accounted by introducing another factor, the relative permeability  $\mu_r$ , giving  $B = \mu_0 \mu_r H$  for magnetic materials.

Thank you